

**5G Toolbox™**

Getting Started Guide



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Phone: 508-647-7000



The MathWorks, Inc.  
1 Apple Hill Drive  
Natick, MA 01760-2098

*5G Toolbox™ Getting Started Guide*

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# Getting Started with 5G Toolbox Software

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## 5G Toolbox Product Description

### **Simulate, analyze, and test the physical layer of 5G communications systems**

5G Toolbox provides standard-compliant functions and reference examples for the modeling, simulation, and verification of 5G communications systems. The toolbox supports link-level simulation, golden reference verification and conformance testing, and test waveform generation.

With the toolbox you can configure, simulate, measure, and analyze end-to-end communications links. You can modify or customize the toolbox functions and use them as reference models for implementing 5G systems and devices.

The toolbox provides reference examples to help you explore baseband specifications and simulate the effects of RF designs and interference sources on system performance. You can generate waveforms and customize test benches to verify that your designs, prototypes, and implementations comply with the 3GPP 5G New Radio (NR) standard.

### **Key Features**

- Standard-compliant models for 3GPP 5G NR Release 15
- Link-level simulation with reference examples, including 5G NR PDSCH throughput simulation
- OFDM waveform generation with NR subcarrier spacings and frame numerologies
- TR 38.901 propagation channel models, including tapped delay line (TDL) and clustered delay line (CDL)
- Downlink transport and physical channels (shared, control, broadcast); synchronization and demodulation reference signals
- Signal processing functions, including channel coding (LDPC and polar codes), channel estimation, synchronization, and equalization
- C and C++ code generation support

# About 5G

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# What Is 5G New Radio?

New Radio (NR) is the air interface supporting the next generation of mobile communication, commonly referred to as fifth generation or 5G.

The predecessors of 5G NR are GSM, UMTS, and LTE, also referred to as second generation (2G), third generation (3G), and fourth generation (4G) technologies, respectively. GSM primarily enabled voice calls. The redesigned interfaces of UMTS and LTE enabled and gradually improved mobile broadband connectivity with high data rates and high efficiency.

5G NR continues on the path of LTE by enabling much higher data rates and much higher efficiency for mobile broadband. However, as a response to the demands of *networked society*, the scope of 5G NR goes beyond mobile broadband connectivity. The main requirement of 5G NR is to enable wireless connectivity everywhere, at any time to anyone and anything.

The wide range of use cases that drive 5G NR are classified by three main scenarios.

- *Enhanced mobile broadband (eMBB)* — This scenario is still the most important usage scenario that addresses human-centric communications. eMBB use cases have various challenges. For example, hot spots require higher data rates, higher user density, and a need for high capacity. Wide area coverage stresses mobility and seamless user experience with lower requirements on data rate and user density.
- *Massive machine type communications (mMTC)* — This scenario addresses pure machine-centric use cases characterized by a large number of connected devices. Typically, the data rate requirement of mMTC applications is low. However, the use cases demand a high connection density locally, low cost, and long battery life.
- *Ultra reliable and low latency communications (URLLC)* — This scenario covers both human-centric communication and critical machine-type communication (C-MTC) that demand low latency, reliability, and high availability. Typical URLLC use cases include 3-D gaming, self driving cars, mission-critical applications, remote medical surgery, and wireless control of industrial equipment.

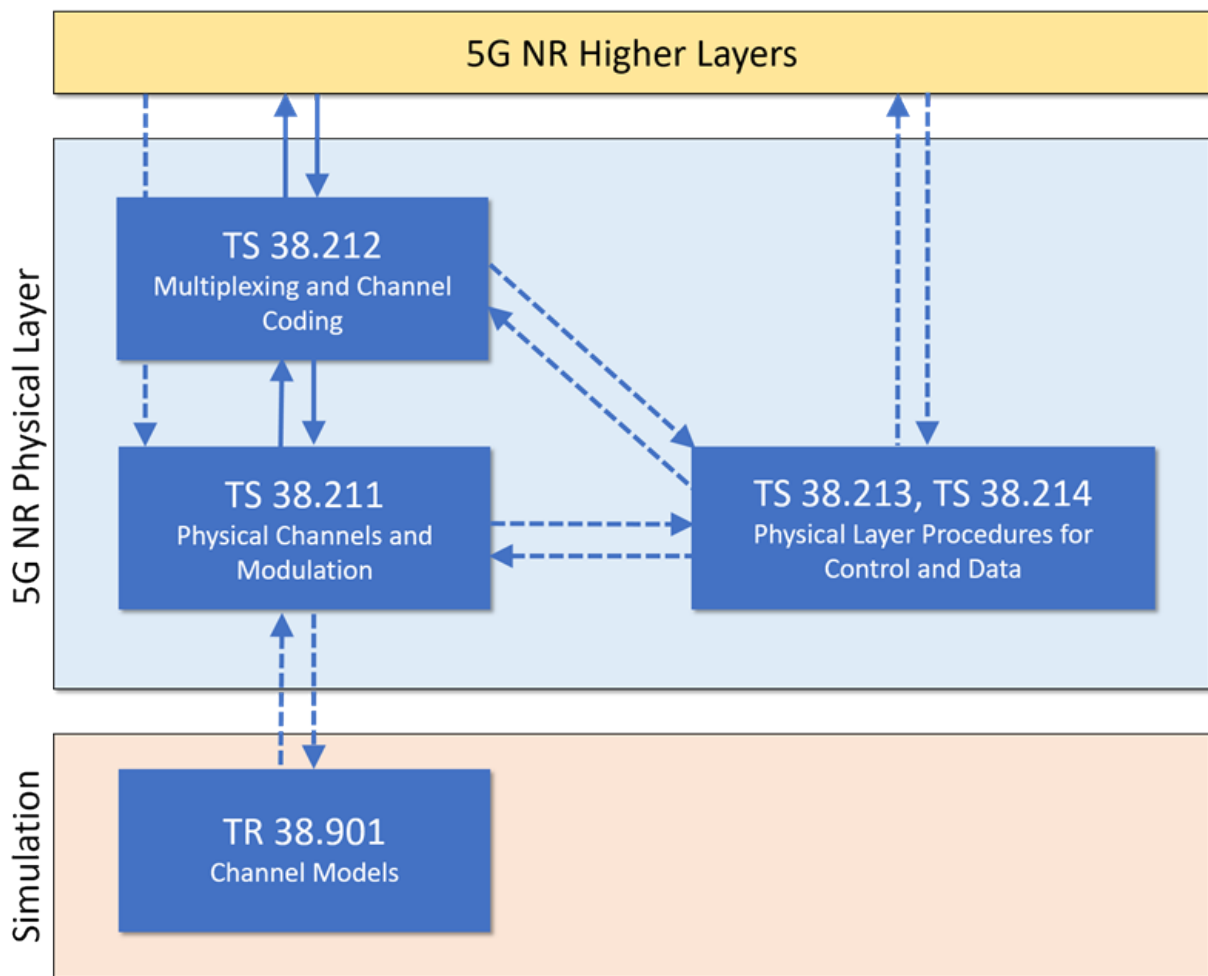
This classification is based on presently foreseen use cases and identifies key capabilities of 5G NR. Based on these capabilities, the 5G NR interface is designed to easily adapt to unforeseen use cases that will evolve and emerge over time.



## Scope of 5G Toolbox

The 5G NR specification is developed by the Third Generation Partnership Project (3GPP). The first release of the standard was frozen in mid-2018 as 3GPP 5G NR Release 15.

5G Toolbox provides implementations for a subset of the 5G NR physical layer specification and channel model specifications. The following diagram highlights the scope of 5G Toolbox in terms of the addressed specifications and their connectivity.



### References

- [1] Dalman, E., S. Parkvall, and J. Sköld. *4G, LTE-Advanced Pro and The Road to 5G*. Kidlington, Oxford: Academic Press, 2016.

### See Also

#### More About

- “5G Toolbox and the 5G NR Protocol Layers” on page 2-5
- “What Is LTE?” (LTE Toolbox)

#### External Websites

- <http://www.3gpp.org>

## 5G Toolbox and the 5G NR Protocol Layers

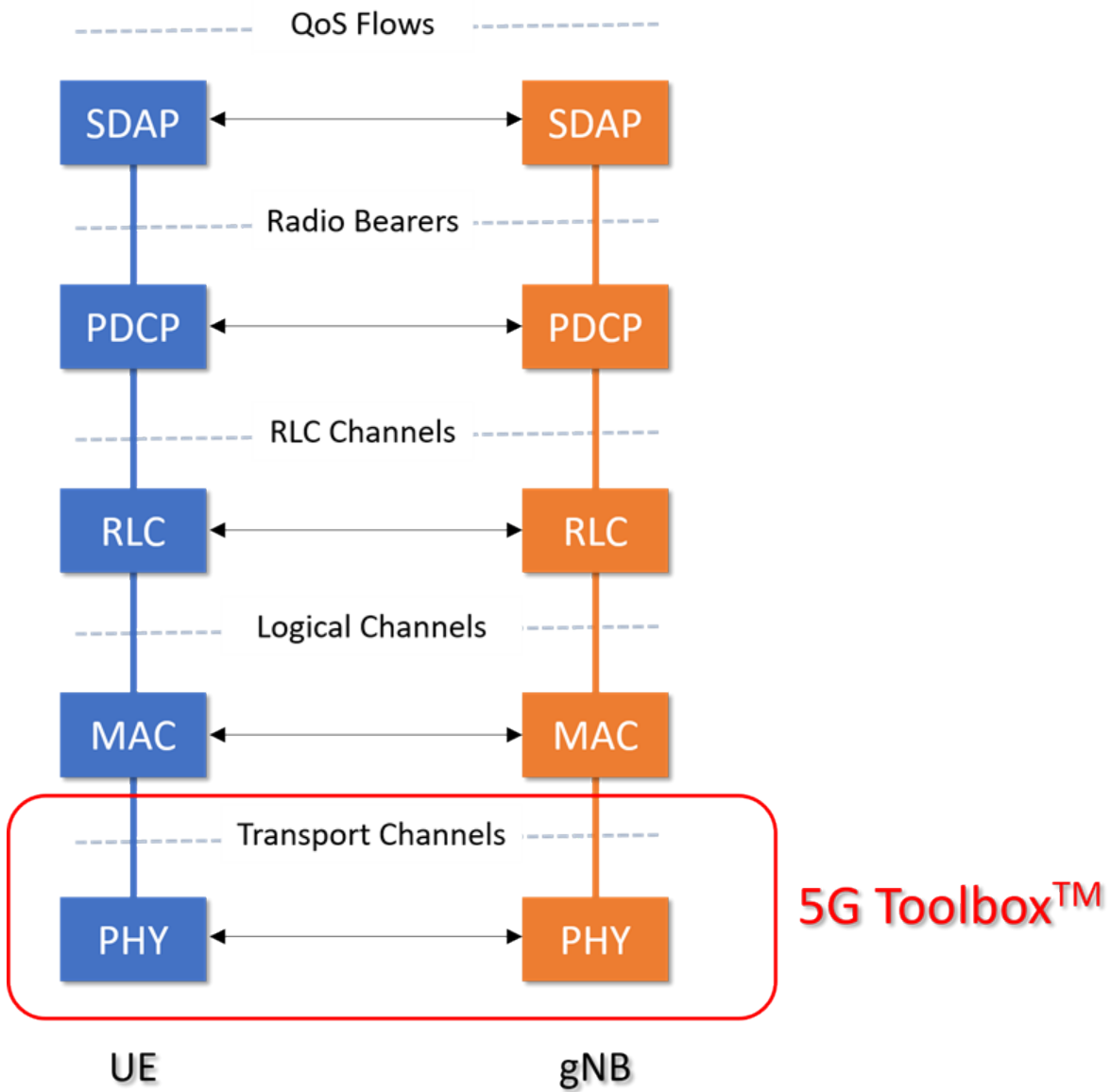
The 5G NR radio access network is comprised of these protocol entities:

- Service data adaptation protocol (SDAP)
- Packet data convergence protocol (PDCP)
- Radio link control (RLC)
- Medium access control (MAC)
- Physical layer (PHY)

The SDAP protocol is new in 5G NR compared to the LTE protocol stack. SDAP handles the new QoS framework of the 5G System (in the 5G Core). SDAP applies also to LTE when connected to the 5G Core. The introduction of SDAP enables end-to-end QoS framework that works in both directions.

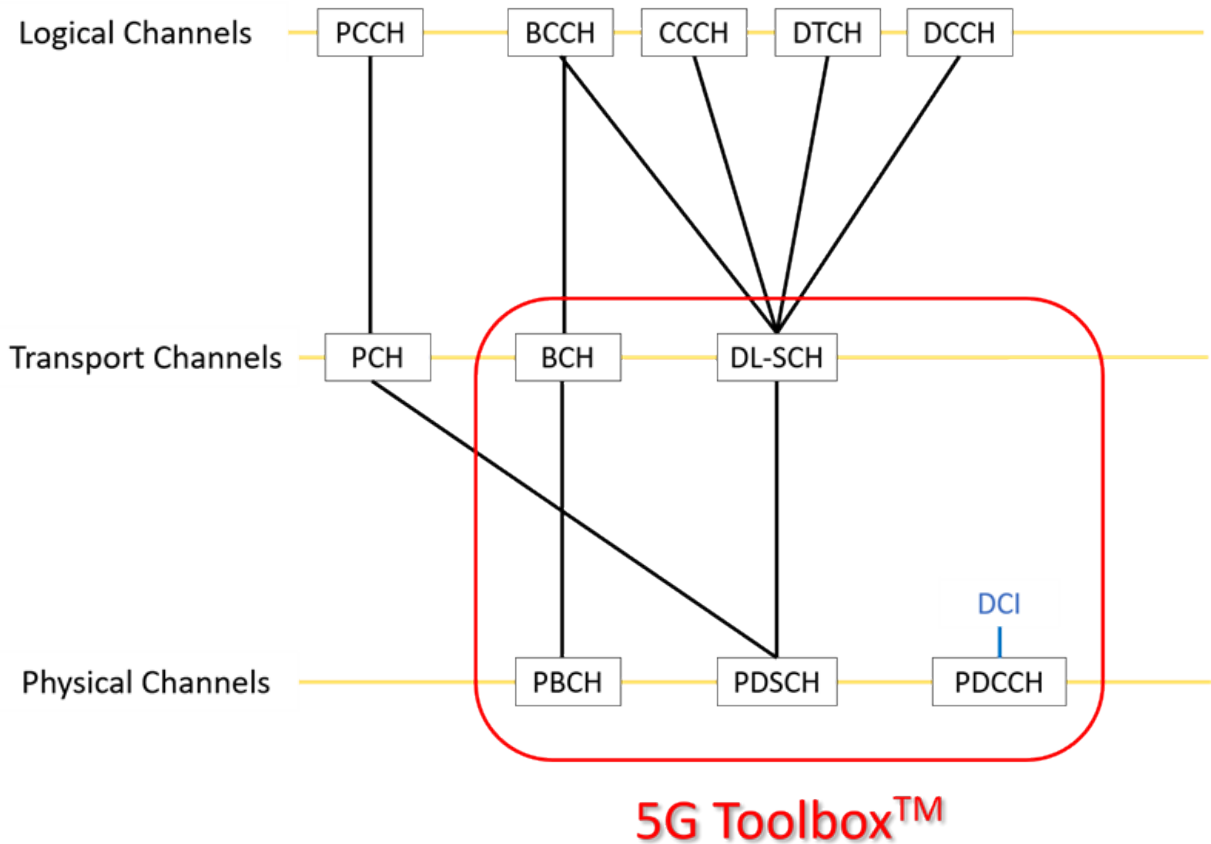
To meet the desired key capabilities of 5G NR, the other layers of the stack provide various enhancements over their LTE counterparts. The PDCP, RLC, and MAC protocols handle tasks such as header compression, ciphering, segmentation and concatenation, and multiplexing and demultiplexing. PHY handles coding and decoding, modulation and demodulation, and antenna mapping.

This figure shows the 5G NR user plane protocol stack for user equipment (UE) and the NR radio access network node (gNB). 5G Toolbox supports the 5G NR physical layer, including physical channels and signals. The toolbox also supports interfacing with portions of the RLC and MAC layers, including transport channels and logical channels.



## Downlink Channel Mapping

5G NR system downlink data follows the mapping between logical channels, transport channels, and physical channels, as indicated in the diagram. 5G Toolbox provides the red-highlighted downlink functionality for physical channels, transport channels, and control information.



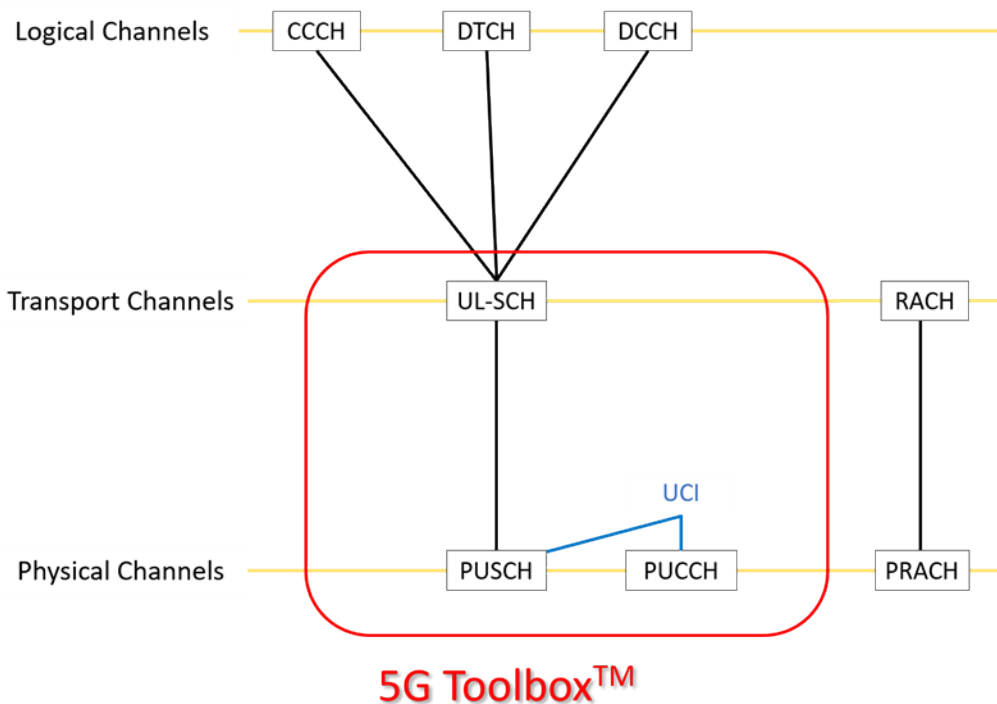
For more details, see “Downlink Channels” or the specific downlink channel:

- “Downlink Physical Signals”
- “Downlink Physical Channels”

- “Downlink Transport Channels”
- “Downlink Control Information”

## Uplink Channel Mapping

5G NR system uplink data follows the mapping between logical channels, transport channels, and physical channels, as indicated in the diagram. 5G Toolbox provides the red-highlighted uplink functionality for physical channels, transport channels, and control information.



For more details, see “Uplink Channels” or the specific uplink channel:

- “Uplink Physical Channels”
- “Uplink Transport Channels”

- “Uplink Control Information”

## References

- [1] 3GPP TS 38.212. “NR; Multiplexing and channel coding.” *3rd Generation Partnership Project; Technical Specification Group Radio Access Network*.
- [2] 3GPP TS 38.300. “NR; NR and NG-RAN Overall Description.” *3rd Generation Partnership Project; Technical Specification Group Radio Access Network*.

## See Also

### More About

- “What Is 5G New Radio?” on page 2-2

### External Websites

- <http://www.3gpp.org>





# Tutorials

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## 5G NR Downlink Carrier Waveform Generation

This example implements a 5G NR downlink carrier waveform generator using 5G Toolbox™.

### Introduction

This example shows how to parameterize and generate a 5G New Radio (NR) downlink waveform. The following channels and signals are generated;

- PDSCH and its associated DM-RS
- PDCCH and its associated DM-RS
- PBCH and its associated DM-RS
- PSS and SSS

This example supports the parameterization and generation of multiple SCS specific carriers and multiple bandwidth parts (BWP). Multiple instances of the PDSCH and PDCCH channels can be generated over the different BWPs. Sets of CORESETs and search space monitoring opportunities can be configured for mapping the PDCCHs. Note that no precoding is applied to the physical channels and signals in this example.

### Waveform and Carrier Configuration

This section sets the SCS specific carrier bandwidths in resource blocks, the cell ID, and the length of the generated waveform in subframes. You can visualize the generated resource grids by setting the `DisplayGrids` field to 1. The channel bandwidth and frequency range parameters are used to display the associated minimum guardbands on a schematic diagram of the SCS carrier alignment.

```
waveconfig = [];  
waveconfig.NCellID = 0;           % Cell identity  
waveconfig.ChannelBandwidth = 40; % Channel bandwidth (MHz)  
waveconfig.FrequencyRange = 'FR1'; % 'FR1' or 'FR2'  
waveconfig.NumSubframes = 10;    % Number of 1ms subframes in generated waveform (1,2)  
waveconfig.DisplayGrids = 1;     % Display the resource grids after signal generation  
  
% Define a set of SCS specific carriers, using the maximum sizes for a  
% 40 MHz NR channel. See TS 38.101-1 for more information on defined  
% bandwidths and guardband requirements  
carriers(1).SubcarrierSpacing = 15;  
carriers(1).NRB = 216;  
carriers(1).RBStart = 0;
```

```
carriers(2).SubcarrierSpacing = 30;
carriers(2).NRB = 106;
carriers(2).RBStart = 1;
```

### SS Burst

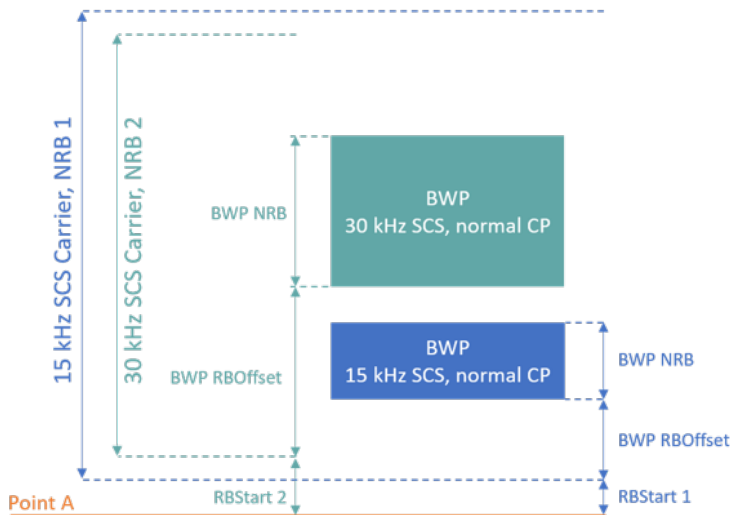
In this section you can set the parameters for the SS burst. The numerology of the SS burst can be different from other parts of the waveform. This is specified via the block pattern parameter as specified in TS 38.213 Section 4.1. A bitmap is used to specify which blocks are transmitted in a 5ms half-frame burst. The periodicity in milliseconds and the power of the burst can also be set here. Other SS burst parameters not shown here can also be set. For the full list see the help for `hSSBurst`.

```
% SS burst configuration
ssburst = [];
ssburst.Enable = 1;
ssburst.BlockPattern = 'Case B';
ssburst.SSBTransmitted = [1 1 1 1];
ssburst.SSBPeriodicity = 20;
ssburst.FrequencySSB = 0*5000;
ssburst.Power = 0;
```

% Enable SS Burst  
 % Case B (30kHz) subcarrier spacing  
 % Bitmap indicating blocks transmitted in a 5ms  
 % SS burst set periodicity in ms (5, 10, 20, 40)  
 % Frequency offset of SS burst (Hz), relative to  
 % Power scaling in dB

### Bandwidth Parts

A BWP is formed by a set of contiguous resources sharing a numerology on a given carrier. This example supports the use of multiple BWPs using a structure array. Each entry in the array represents a BWP. For each BWP you can specify the subcarrier spacing (SCS), the cyclic prefix (CP) length and the bandwidth. The `SubcarrierSpacing` parameter maps the BWP to one of the SCS specific carriers defined earlier. The `RBOffset` parameter controls the location of the BWP in the carrier. This is expressed in terms of the BWP numerology. Different BWPs can overlap with each other.



% Bandwidth parts configurations

```
bwp = [];
```

```
bwp(1).SubcarrierSpacing = 15;
bwp(1).CyclicPrefix = 'Normal';
bwp(1).NRB = 25;
bwp(1).RBOffset = 10;
```

```
% BWP Subcarrier Spacing
% BWP Cyclic prefix for 15 kHz
% Size of BWP
% Position of BWP in SCS carrier
```

```
bwp(2).SubcarrierSpacing = 30;
bwp(2).CyclicPrefix = 'Normal';
bwp(2).NRB = 50;
bwp(2).RBOffset = 50;
```

```
% BWP Subcarrier Spacing
% BWP Cyclic prefix for 30 kHz
% Size of BWP
% Position of BWP in SCS carrier
```

### CORESET and Search Space Configuration

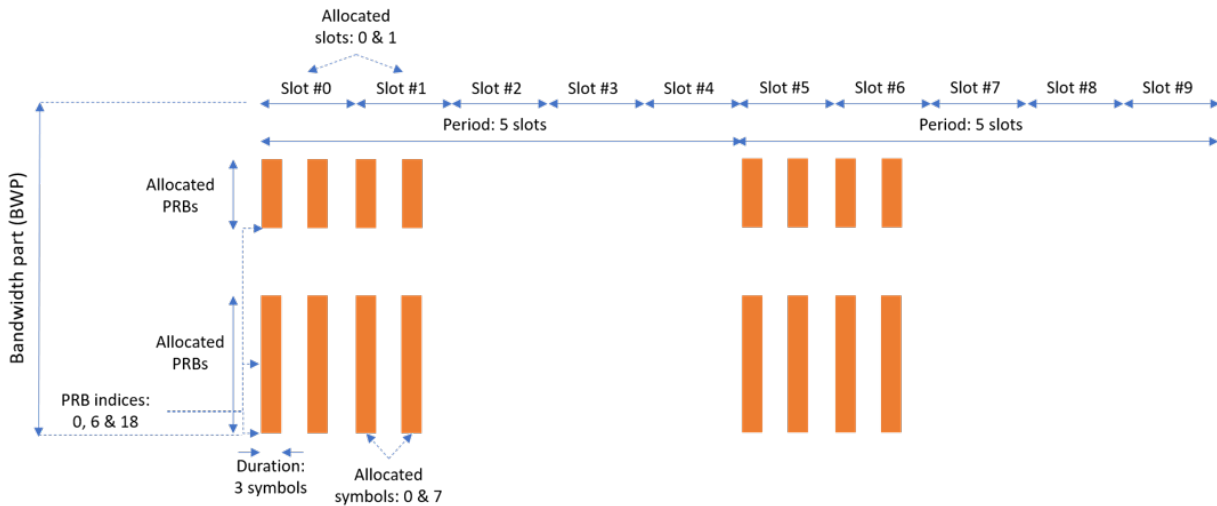
The parameters in this section specify the control resource set (CORESET) and the PDCCH search space configuration. The CORESET specifies the possible locations (in time and frequency) of the control channel for a given numerology. This example supports multiple CORESETs. The following parameters can be specified:

- Allocated OFDM symbols: specifies the first symbol of each CORESET monitoring opportunity in a slot
- The allocated slots within a period

- Periodicity of the allocation. If this is set to empty it indicates no repetition
- CORESET duration in symbols, either 1, 2 or 3.
- The first PRB of the allocation. Note that the allocation is in blocks of 6 PRBs.

Note that this example only supports non-interleaved CCE-to-REG mapped CORESETs.

The figure below shows the meaning of the CORESET parameters.



#### % CORESET/search configurations

```
coreset = [];
coreset(1).AllocatedSymbols = [0,7];
coreset(1).AllocatedSlots = [0,1];
coreset(1).AllocatedPeriod = 5;
coreset(1).Duration = 3;
coreset(1).AllocatedPRB = 6*[0,1,3];
```

```
% First symbol of each CORESET monitoring oppo
% Allocated slots within a period
% Allocated slot period (empty implies no repe
% CORESET symbol duration (1,2,3)
% 6 REG sized indices, relative to BWP (RRC -
```

### PDCCH Instances Configuration

This section specifies the parameters for the set of PDCCH instances in the waveform. Each element in the structure array defines a PDCCH sequence instance. The following parameters can be set:

- Enable/disable the PDCCH sequence

- Specify the BWP carrying the PDCCH
- PDCCH instance power in dB
- Enable/disable DCI channel coding
- Allocated search spaces: indices within the corresponding CORESET where to map the PDCCH
- CORESET which carries the PDCCH instance
- Periodicity of the allocation. If this is set to empty it indicates no repetition
- Number of control channel elements (CCEs) in this PDCCH
- NumCCE and StartCCE specify the elements used for the transmission of this PDCCH
- RNTI
- Scrambling NID for this PDCCH and its associated DM-RS
- DM-RS power boosting
- DCI message payload size
- DCI message data source. You can use one of the following standard PN sequences: 'PN9-ITU', 'PN9', 'PN11', 'PN15', 'PN23'. The seed for the generator can be specified using a cell array in the form {'PN9', seed}. If no seed is specified, the generator is initialized with all ones

```

pdcch = [];
pdcch(1).Enable = 1;           % Enable PDCCH sequence
pdcch(1).BWP = 1;             % Bandwidth part
pdcch(1).Power = 1.1;        % Power scaling in dB
pdcch(1).EnableCoding = 1;   % Enable DCI coding
pdcch(1).AllocatedSearchSpaces = [0,3]; % Index within the CORESET
pdcch(1).CORESET = 1;        % Control resource set ID which carries this PDCCH
pdcch(1).AllocatedPeriod = 4; % Allocation slot period (empty implies no repetition)
pdcch(1).NumCCE = 8;         % Number of CCE used by PDCCH
pdcch(1).StartCCE = 0;       % Starting CCE of PDCCH
pdcch(1).RNTI = 0;           % RNTI
pdcch(1).NID = 1;            % PDCCH and DM-RS scrambling NID
pdcch(1).PowerDMRS = 0;      % Additional power boosting in dB
pdcch(1).DataBlkSize = 20;   % DCI payload size
pdcch(1).DataSource = 'PN9'; % DCI data source

```

### PDSCH Instances Configuration

This section specifies the set of PDSCH instances in the waveform. Each element in the structure array defines a PDSCH sequence instance. This example defines two PDSCH sequence instances.

## General Parameters

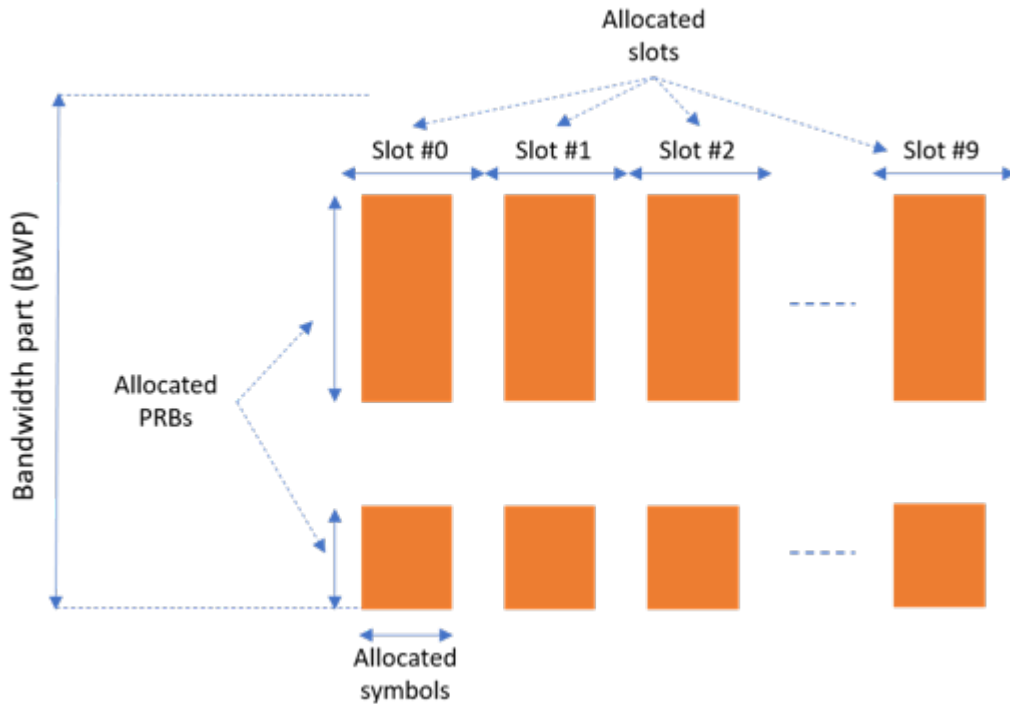
The following parameters are set for each instance:

- Enable/disable this PDSCH sequence
- Specify the BWP carrying the PDSCH. The PDSCH will use the SCS specified for this BWP
- Power scaling in dB
- Enable/disable DL-SCH transport channel coding
- Transport block data source. You can use one of the following standard PN sequences: 'PN9-ITU', 'PN9', 'PN11', 'PN15', 'PN23'. The seed for the generator can be specified using a cell array in the form { 'PN9' , seed}. If no seed is specified, the generator is initialized with all ones
- Target code rate used to calculate the transport block sizes
- Overhead parameter
- Symbol modulation
- Number of layers
- Redundancy version (RV) sequence

```
pdsch = [];
pdsch(1).Enable = 1;           % Enable PDSCH sequence
pdsch(1).BWP = 1;             % Bandwidth part
pdsch(1).Power = 0;          % Power scaling in dB
pdsch(1).EnableCoding = 1;    % Enable DL-SCH transport channel coding
pdsch(1).DataSource = 'PN9'; % Channel data source
pdsch(1).TargetCodeRate = 0.4785; % Code rate used to calculate transport block s
pdsch(1).Xoh_PDSCH = 0;      % Rate matching overhead
pdsch(1).Modulation = 'QPSK'; % 'QPSK', '16QAM', '64QAM', '256QAM'
pdsch(1).NLayers = 2;        % Number of PDSCH layers
pdsch(1).RVSequence = [0,2,3,1]; % RV sequence to be applied cyclically across t
```

## Allocation

The following diagram represents some of the parameters used in the PDSCH allocation.



You can set the following parameters to control the PDSCH allocation. Note that these parameters are relative to the BWP. The specified PDSCH allocation will avoid the locations used for the SS burst.

- Symbols in a slot allocated to each PDSCH instance
- Slots in a frame used for the sequence of PDSCH
- Period of the allocation in slots. If this is empty it indicates no repetition
- The allocated PRBs are relative to the BWP
- RNTI. This value is used to link the PDSCH to an instance of the PDCCH
- NID for scrambling the PDSCH bits

```

pdsch(1).AllocatedSymbols = 2:10;           % Range of symbols in a slot
pdsch(1).AllocatedSlots = [0:9];          % Allocated slot indices for PDSCH sequence
pdsch(1).AllocatedPeriod = 15;            % Allocation period in slots (empty implies no repetition)
pdsch(1).AllocatedPRB = [0:5, 10:20];    % PRB allocation
    
```



```
pdsch(1).RNTI = 0;           % RNTI
pdsch(1).NID = 1;          % Scrambling for data part
```

Note that the generator in this example does not check for inter-channel conflict. However, additional parameters can be specified for rate matching around other resources, if required

- The PDSCH can be rate matched around one or more CORESETs
- The PDSCH can be rate matched around other resource allocations

```
pdsch(1).RateMatch(1).CORESET = [1];           % Rate matching pattern, defined
pdsch(1).RateMatch(1).Pattern.AllocatedPRB = []; % Rate matching pattern, defined
pdsch(1).RateMatch(1).Pattern.AllocatedSymbols = [];
pdsch(1).RateMatch(1).Pattern.AllocatedSlots = [];
pdsch(1).RateMatch(1).Pattern.AllocatedPeriod = [];
```

### PDSCH DM-RS Configuration

The following DM-RS parameters can be set

```
% Antenna port and DM-RS configuration (TS 38.211 section 7.4.1.1)
pdsch(1).PortSet = 0:pdsch(1).NLayers-1; % DM-RS antenna ports used
pdsch(1).PDSCHMappingType = 'A';         % PDSCH mapping type ('A'(slot-wise),'B'(non s
pdsch(1).DMRSTypeAPosition = 2;          % Mapping type A only. First DM-RS symbol pos
pdsch(1).DMRSLength = 1;                 % Number of front-loaded DM-RS symbols (1(sing
pdsch(1).DMRSAdditionalPosition = 0;      % Additional DM-RS symbol positions (max rang
pdsch(1).DMRSConfigurationType = 2;      % DM-RS configuration type (1,2)
pdsch(1).NumCDMGroupsWithoutData = 0;    % CDM groups without data (max range 0...3)
pdsch(1).NIDNSCID = 1;                   % Scrambling identity (0...65535)
pdsch(1).NSCID = 0;                       % Scrambling initialization (0,1)
pdsch(1).PowerDMRS = 0;                   % Additional power boosting in dB
```

### Specifying Multiple PDSCH Instances

A second PDSCH sequence instance is specified next using the second BWP.

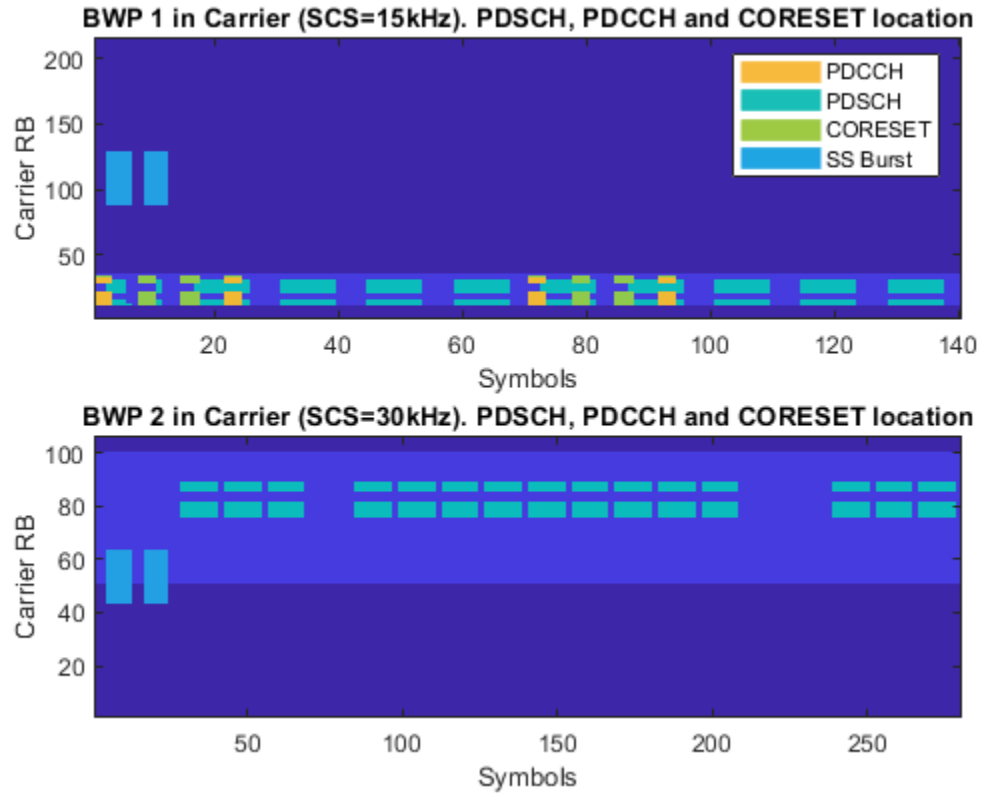
```
pdsch(2) = pdsch(1);
pdsch(2).Enable = 1;
pdsch(2).BWP = 2; % PDSCH mapped to 2nd BWP
pdsch(2).AllocatedSymbols = 0:11;
pdsch(2).AllocatedSlots = [2:4,6:20];
pdsch(2).AllocatedPRB = [25:30, 35:38]; % PRB allocation, relative to BWP
```

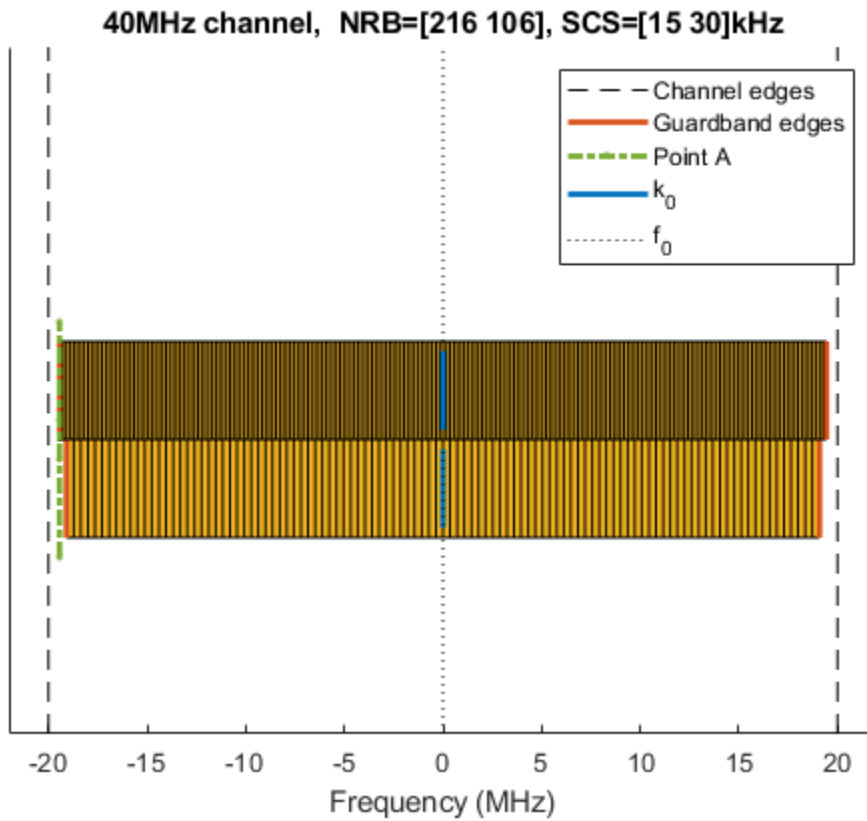
### Waveform Generation

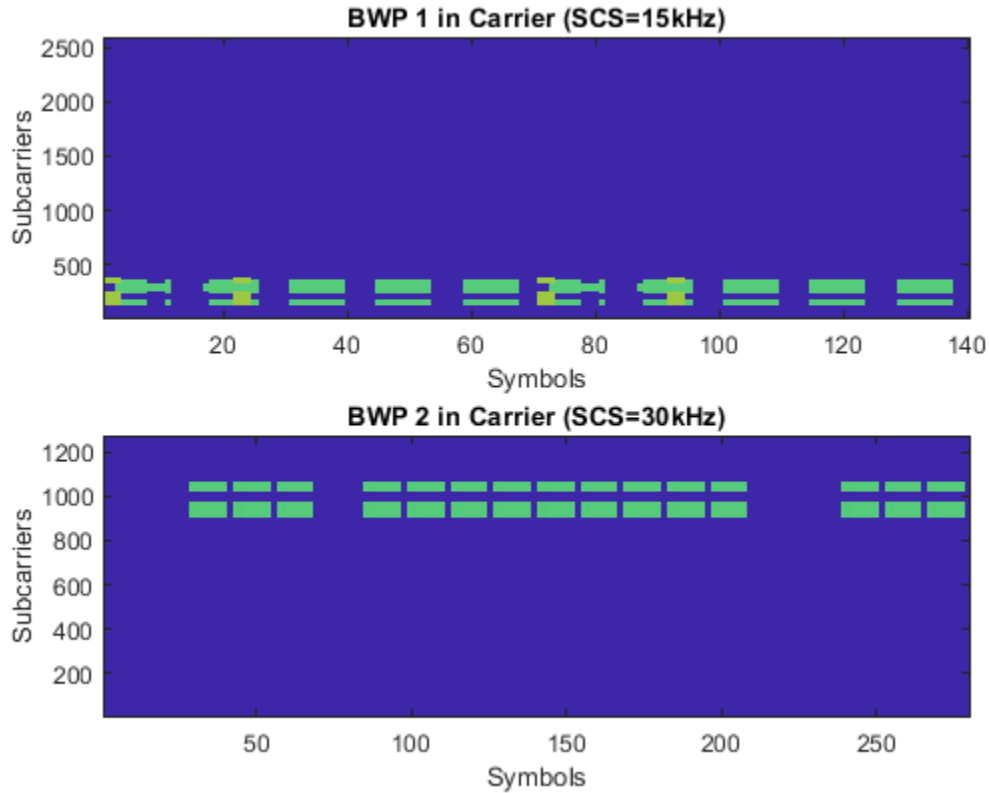
This section collects all the parameters into the carrier configuration and generates the waveform.

```
% Collect together channel oriented parameter sets into a single
% configuration
waveconfig.SSBurst = ssburst;
waveconfig.Carriers = carriers;
waveconfig.BWP = bwp;
waveconfig.CORESET = coreset;
waveconfig.PDCCH = pdcch;
waveconfig.PDSCH = pdsch;

% Generate complex baseband waveform
[waveform,bwpset] = hNRDownlinkWaveformGenerator(waveconfig);
```







The waveform generator also plots the resource grids for the bandwidth parts (this is controlled by the field `DisplayGrids` in the carrier configuration). The following plots are generated:

- The SCS specific carrier resource grids, along with the minimum guardbands, aligned relative to the overall channel bandwidth.
- Resource grids showing the location of the components (PDCCH, PDSCH, and CORESET) in each BWP. This does not plot the power of the signals, just their location in the grid.
- Generated waveform in the frequency domain for each BWP. This includes the PDCCH and PDSCH instances.

Note that none of these resource grids include the SS burst which is independent of the BWPs.

The waveform generator function returns the time domain waveform and a structure array `bwpset`, which contains the following fields:

- The resource grid corresponding to this BWP
- The resource grid of the overall bandwidth containing the channels and signals in this BWP
- An info structure with information corresponding to the BWP. The contents of this info structure for the first BWP are shown below.

```
disp('Information associated to BWP 1:')  
disp(bwpset(1).Info)
```

```
Information associated to BWP 1:  
    SamplingRate: 61440000  
        Nfft: 4096  
    Windowing: 10  
CyclicPrefixLengths: [1×14 double]  
    SymbolLengths: [1×14 double]  
    NSubcarriers: 2592  
    SubcarrierSpacing: 15  
    SymbolsPerSlot: 14  
    SlotsPerSubframe: 1  
    SymbolsPerSubframe: 14  
    SamplesPerSubframe: 61440  
    SubframePeriod: 1.0000e-03  
    Midpoints: [1×141 double]  
    WindowOverlap: [10 10 10 10 10 10 10 10 10 10 10 10 10 10]  
        k0: 0
```

Note that the generated resource grid is a 3D matrix where the different planes represent the antenna ports. For the different physical channels and signals the lowest port is mapped to the first plane of the grid.

## See Also

### Functions

`nrPBCH` | `nrPBCHDMRS` | `nrPDCCH` | `nrPDSCH` | `nrPSS` | `nrSSS`

## **More About**

- “5G NR Uplink Carrier Waveform Generation” on page 3-16

## 5G NR Uplink Carrier Waveform Generation

This example implements a 5G NR uplink carrier waveform generator using 5G Toolbox™.

### Introduction

This example shows how to parameterize and generate a 5G New Radio (NR) uplink waveform. The following channels and signals are generated

- PUSCH and its associated DM-RS
- PUCCH and its associated DM-RS

This example supports the parameterization and generation of multiple bandwidth parts (BWP). Multiple instances of the PUSCH and PUCCH channels can be generated over the different BWPs. The example allows to configure PUCCH and PUSCH for a specific UE categorized by RNTI and transmits only PUSCH for that specific RNTI when both PUCCH and PUSCH overlap in a slot.

### Waveform and Carrier Configuration

This section sets the subcarrier spacing (SCS) specific carrier bandwidths in resource blocks, the physical layer cell identity NCellID, and the length of the generated waveform in subframes. You can visualize the generated resource grids by setting the DisplayGrids field to 1. The channel bandwidth and frequency range parameters are used to display the associated minimum guardbands on a schematic diagram of the SCS carrier alignment. The schematic diagram is displayed in one of the output plots of the example.

```
waveconfig = [];  
waveconfig.NCellID = 0;           % Cell identity  
waveconfig.ChannelBandwidth = 50; % Channel bandwidth (MHz)  
waveconfig.FrequencyRange = 'FR1'; % 'FR1' or 'FR2'  
waveconfig.NumSubframes = 10;    % Number of 1ms subframes in generated waveform  
                                  % (1,2,4,8 slots per 1ms subframe,  
                                  % depending on SCS)  
waveconfig.DisplayGrids = 1;     % Display the resource grids after signal generation  
  
% Define a set of SCS specific carriers, using the maximum sizes for a 50  
% MHz NR channel. See TS 38.101-1 for more information on defined  
% bandwidths  
carriers = [];  
carriers(1).SubcarrierSpacing = 15;
```

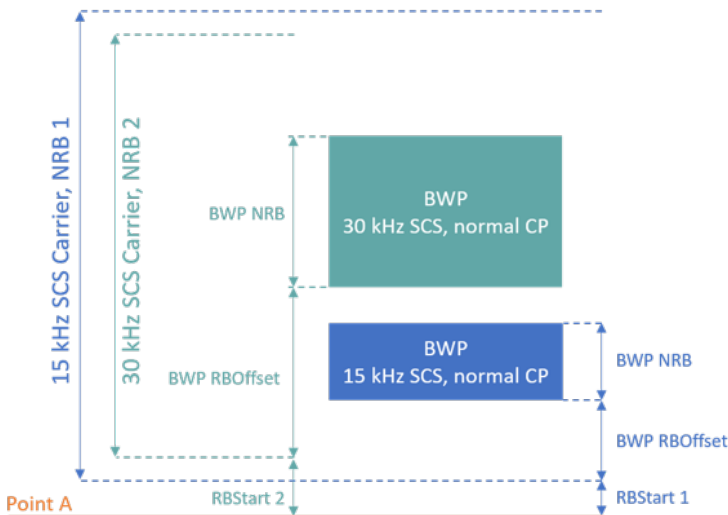


```
carriers(1).NRB = 270;
carriers(1).RBStart = 0;
```

```
carriers(2).SubcarrierSpacing = 30;
carriers(2).NRB = 133;
carriers(2).RBStart = 1;
```

### Bandwidth Parts

A BWP is formed by a set of contiguous resources sharing a numerology on a given SCS specific carrier. This example supports the use of multiple BWPs using a struct array. Each entry in the array represents a BWP. For each BWP you can specify the subcarrier spacing (SCS), the cyclic prefix (CP) length and the bandwidth. The SubcarrierSpacing parameter maps the BWP to one of the SCS specific carriers defined earlier. The RBOffset parameter controls the location of the BWP in the carrier. This is expressed in terms of the BWP numerology. Different BWPs can overlap with each other.



```
% Bandwidth parts configurations
```

```
bwp = [];
```

```
bwp(1).SubcarrierSpacing = 15;
bwp(1).CyclicPrefix = 'Normal';
bwp(1).NRB = 25;
```

```
% BWP1 Subcarrier Spacing
% BWP1 cyclic prefix
% Size of BWP1
```

```

bwp(1).RBOffset = 10; % Position of BWP1 in carrier

bwp(2).SubcarrierSpacing = 30; % BWP2 Subcarrier Spacing
bwp(2).CyclicPrefix = 'Normal'; % BWP2 cyclic prefix
bwp(2).NRB = 51; % Size of BWP2
bwp(2).RBOffset = 40; % Position of BWP2 in carrier

```

#### **PUCCH Instances Configuration**

This section specifies the parameters for the set of PUCCH instances in the waveform. Each element in the structure array defines a PUCCH sequence instance. The following parameters can be set:

- Enable/disable the PUCCH sequence
- Specify the BWP carrying the PUCCH
- PUCCH instance power in dB
- Slots within a period used for PUCCH
- Periodicity of the allocation. Use empty to indicate no repetition
- DM-RS power boosting in dB

```

pucch = [];
pucch(1).Enable = 1; % Enable PUCCH sequence
pucch(1).BWP = 1; % Bandwidth part
pucch(1).Power = 0; % Power scaling in dB
pucch(1).AllocatedSlots = [3 4]; % Allocated slots within a period
pucch(1).AllocatedPeriod = 6; % Allocation slot period (empty implies no
pucch(1).PowerDMRS = 1; % Additional power boosting in dB

```

#### **PUCCH Resource Configuration**

This section specifies the PUCCH sequence resource related parameters. The parameters can be categorized into the following sections:

- Enable/Disable the PUCCH dedicated resource. If this is enabled, it uses dedicated resource as per TS 38.213 Section 9.2.1
- Provide the resource index value (0...15), when dedicated resource is enabled

When the dedicated resource is disabled, the resource index needs to be provided. The resource and format parameters for the PUCCH transmission are filled up directly based on the resource index. All the other parameters that are provided for resource and format configurations are not considered.

When the dedicated resource is enabled, the following resource parameters need to be provided:

- Specify the index of first PRB prior to frequency hopping or for no frequency hopping within the BWP
- Specify the index of first PRB after frequency hopping within the BWP
- Intra-slot frequency hopping configuration ('enabled', 'disabled')
- Group hopping configuration ('neither', 'enable', 'disable')

and the following format specific parameters need to be provided

- PUCCH format configuration in the resource (0..4)
- Starting symbol index allocated for PUCCH transmission
- Number of OFDM symbols allocated for PUCCH transmission. For PUCCH formats 1, 3 and 4, the number of OFDM symbols allocated are in range 4 to 14, and for formats 0 and 2, it is either 1 or 2
- Initial cyclic shift for formats 0 and 1. The value is in range 0 to 11
- Modulation scheme for formats 3 and 4 ('QPSK', 'pi/2-BPSK')
- Number of resource blocks allocated for format 2 and 3. The nominal value is one of the set {1,2,3,4,5,6,8,9,10,12,15,16}
- Spreading factor for format 4. The value is either 2 or 4
- Orthogonal cover code index for formats 1 and 4. For format 1, the value is in range 0 to 6. For format 4, the value is in between 0 and less than the spreading factor
- Indicate the presence of additional DM-RS for formats 3 and 4. The value is either 0 or 1

Scrambling identities to be used for different formats

- RNTI for formats 2/3/4. It is used for sequence generation. It is in range 0 to 65535
- Scrambling identity (NID) for PUCCH formats 2/3/4. It is in range 0 to 1023. Use empty ([]) to use physical layer cell identity. It is used in sequence generation. This parameter is provided by higher-layer parameter *dataScramblingIdentityPUSCH*
- PUCCH hopping identity for formats 0/1/3/4. Use empty ([]) to use physical layer cell identity. The value is used in sequence generation for format 0, both sequence and DM-RS generation for format 1 and only for DM-RS generation for formats 3 and 4
- DM-RS scrambling NID for PUCCH format 2. It is in range 0 to 65535. Use empty ([]) to use physical layer cell identity

Irrespective of dedicated resource configuration, the following parameters are to be provided for slot repetitions:

- Specify the number of slot repetitions for formats 1,3,4 (2 or 4 or 8). For no slot repetition, the value can be specified as 1
- Specify the inter-slot frequency hopping for formats 1,3,4 ('enabled','disabled'). If this is enabled and the number of slot repetitions is more than one, then intra-slot frequency hopping is disabled
- Specify the maximum code rate. The nominal value is one of the set {0.08, 0.15, 0.25, 0.35, 0.45, 0.6, 0.8}

```

% Dedicated resource parameters
pucch(1).DedicatedResource = 1;           % Enable/disable the dedicated resource con
% Provide the resource index value when dedicated resource is disabled. The
% PUCCH resource is configured based on the resource index value, as per
% the table 9.2.1-1 of Section 9.2.1, TS 38.213. The resource index value
% is ignored when dedicated resource is enabled or when the dedicated
% resource is disabled and cyclic prefix of PUCCH transmission in the BWP
% is extended.
pucch(1).ResourceIndex = 0;               % Resource index for PUCCH dedicated resour

% Resource parameters (used when dedicated resource is disabled)
pucch(1).StartPRB = 0;                    % Index of first PRB prior to frequency hop
pucch(1).SecondHopPRB = 1;                % Index of first PRB after frequency hoppin
pucch(1).IntraSlotFreqHopping = 'enabled'; % Indication for intra-slot frequency hopp
pucch(1).GroupHopping = 'enable';         % Group hopping configuration ('enable','d

% Format specific parameters
pucch(1).PUCCHFormat = 3;                 % PUCCH format 0/1/2/3/4
pucch(1).StartSymbol = 3;                 % Starting symbol index
pucch(1).NrOfSymbols = 11;                % Number of OFDM symbols allocated for PUC
pucch(1).InitialCS = 3;                   % Initial cyclic shift for format 0 and 1
pucch(1).OCCI = 0;                        % Orthogonal cover code index for format 1
pucch(1).Modulation = 'QPSK';              % Modulation for format 3/4 ('pi/2-BPSK','
pucch(1).NrOfRB = 9;                      % Number of resource blocks for format 2/3
pucch(1).SpreadingFactor = 4;              % Spreading factor for format 4, value is e
pucch(1).AdditionalDMRS = 1;               % Additional DM-RS (0/1) for format 3/4

% Scrambling identities of PUCCH and PUCCH DMRS
pucch(1).RNTI = 0;                        % RNTI (0...65535) for formats 2/3/4
pucch(1).NID = 1;                         % PUCCH scrambling identity (0...1023) for
pucch(1).HoppingId = 1;                   % PUCCH hopping identity (0...1023) for fo
pucch(1).NIDDMRS = 1;                     % DM-RS scrambling identity (0...65535) for

```

```

% Multi-slot configuration parameters
pucch(1).NrOfSlots = 1; % Number of slots for PUCCH repetition (1/2)
pucch(1).InterSlotFreqHopping = 'disabled'; % Indication for inter-slot frequency hopping

% Code rate - This parameter is used when there is multiplexing of UCI part
% 1 (HARQ-ACK, SR, CSI part 1) and UCI part 2 (CSI part 2) to get the rate
% matching lengths of each UCI part
pucch(1).MaxCodeRate = 0.15; % Maximum code rate (0.08, 0.15, 0.25, 0.3)

```

### UCI payload configuration

Configure the UCI payload based on the format configuration

- Enable or disable the UCI coding for formats 2/3/4
- Number of HARQ-ACK bits. For formats 0 and 1, value can be at most 2. Set the value to 0, for no HARQ-ACK transmission
- Number of SR bits. For formats 0 and 1, value can be at most 1. Set the value to 0, for no SR transmission
- Number of CSI part 1 bits for formats 2/3/4. Set value to 0, for no CSI part 1 transmission
- Number of CSI part 2 bits for formats 3/4. Set value to 0, for no CSI part 2 transmission. The value is ignored when there are no CSI part 1 bits

Note that the generator in the example transmits UCI information on PUSCH whenever there is an overlap between PUCCH and PUSCH for a specific RNTI in a BWP. The parameters to be configured for UCI transmission on PUSCH are provided in the section UCI on PUSCH. It requires the lengths of UCI and UL-SCH to be transmitted on PUSCH.

```

pucch(1).EnableCoding = 1; % Enable UCI coding
pucch(1).LenACK = 5; % Number of HARQ-ACK bits
pucch(1).LenSR = 5; % Number of SR bits
pucch(1).LenCSI1 = 10; % Number of CSI part 1 bits (for formats 2, 3)
pucch(1).LenCSI2 = 10; % Number of CSI part 2 bits (for formats 3, 4)

pucch(1).DataSource = 'PN9'; % UCI data source

```

```

% UCI message data source. You can use one of the following standard PN
% sequences: 'PN9-ITU', 'PN9', 'PN11', 'PN15', 'PN23'. The seed for the
% generator can be specified using a cell array in the form |{'PN9',seed}|.
% If no seed is specified, the generator is initialized with all ones

```

### Specifying Multiple PUCCH Instances

A second PUCCH sequence instance is specified next using the second BWP.

```
% PUCCH sequence instance specific to second BWP
pucch(2) = pucch(1);
pucch(2).BWP = 2;
pucch(2).StartSymbol = 10;
pucch(2).NrOfSymbols = 2;
pucch(2).PUCCHFormat = 2;
pucch(2).AllocatedSlots = 0:2;
pucch(2).AllocatedPeriod = [];
pucch(2).RNTI = 10;
```

### **PUSCH Instances Configuration**

This section specifies the set of PUSCH instances in the waveform using a struct array. This example defines two PUSCH sequence instances.

#### **General Parameters**

The following parameters are set for each instance:

- Enable/disable this PUSCH sequence
- Specify the BWP this PUSCH maps to. The PUSCH will use the SCS specified for this BWP
- Power scaling in dB
- Enable/disable the UL-SCH transport coding
- Scrambling identity (NID) for PUSCH bits. It is in range 0 to 1023. Use empty ([]) to use physical layer cell identity
- RNTI
- Transform precoding (0,1). The value of 1, enables the transform precoding and the resultant waveform is DFT-s-OFDM. When the value is 0, the resultant waveform is CP-OFDM
- Target code rate used to calculate the transport block sizes.
- Overhead parameter. It is used to calculate the length of transport block size. It is one of the set {0, 6, 12, 18}
- Transmission scheme ('codebook','nonCodebook'). When the transmission scheme is 'codebook', the MIMO precoding is enabled and a precoding matrix is selected based on the number of layers, number of antenna ports and the transmitted precoding matrix indicator. When the transmission is set to 'nonCodebook', an identity matrix is used, leading to no MIMO precoding

- Modulation scheme ('pi/2-BPSK', 'QPSK', '16QAM', '64QAM', '256QAM'). Nominally, the modulation scheme 'pi/2-BPSK' is used when transform precoding is enabled
- Number of layers (1...4). The number of layers is restricted to a maximum of 4 in uplink as there is only one code word transmission. Nominally, the number of layers is set to 1 when transform precoding is enabled
- Number of antenna ports (1,2,4). It is used when codebook transmission is enabled. The number of antenna ports must be greater than or equal to number of layers
- Transmitted precoding matrix indicator (0...27). It depends on the number of layers and the number of antenna ports
- Redundancy version (RV) sequence
- Intra-slot frequency hopping ('enabled','disabled')
- Resource block offset for second hop. It is used when frequency (Intra-slot/Inter-slot) hopping is enabled
- Inter-slot frequency hopping ('enabled','disabled'). If this is enabled, intra-slot frequency hopping is disabled, the starting position of resource block in the allocated PRB of PUSCH in the bandwidth part depends on the whether the slot is even-numbered or odd-numbered
- Transport block data source. You can use one of the following standard PN sequences: 'PN9-ITU', 'PN9', 'PN11', 'PN15', 'PN23'. The seed for the generator can be specified using a cell array in the form { 'PN9' , seed}. If no seed is specified, the generator is initialized with all ones

```

pusch = [];
pusch(1).Enable = 1;           % Enable PUSCH config
pusch(1).BWP = 1;             % Bandwidth part
pusch(1).Power = 0;          % Power scaling in dB
pusch(1).EnableCoding = 1;    % Enable the UL-SCH transport coding
pusch(1).NID = 1;            % Scrambling for data part (0...1023)
pusch(1).RNTI = 0;           % RNTI
pusch(1).TransformPrecoding = 0; % Transform precoding flag (0 or 1)
pusch(1).TargetCodeRate = 0.47; % Code rate used to calculate transport block size
pusch(1).Xoh_PUSCH = 0;      % Overhead. It is one of the set {0,6,12,18}

% Transmission settings
pusch(1).TxScheme = 'codebook'; % Transmission scheme ('codebook','nonCodebook')
pusch(1).Modulation = 'QPSK'; % 'pi/2-BPSK','QPSK','16QAM','64QAM','256QAM'
pusch(1).NLayers = 2; % Number of PUSCH layers (1...4)
pusch(1).NAntennaPorts = 4; % Number of antenna ports (1,2,4). It must be greater than or equal to number of layers
pusch(1).TPMI = 0; % Transmitted precoding matrix indicator (0...27)
pusch(1).RVSequence = [0 2 3 1]; % RV sequence to be applied cyclically across layers

```

```

pusch(1).IntraSlotFreqHopping = 'disabled'; % Intra-slot frequency hopping ('enabled',
pusch(1).RBOffset = 10; % Resource block offset for second hop

% Multi-slot transmission
pusch(1).InterSlotFreqHopping = 'enabled'; % Inter-slot frequency hopping ('enabled',

% Data source
pusch(1).DataSource = 'PN9'; % Transport block data source

```

### Allocation

You can set the following parameters to control the PUSCH allocation.

- PUSCH mapping type. It can be either 'A' or 'B'.
- Symbols in a slot where the PUSCH is mapped to. It needs to be a contiguous allocation. For PUSCH mapping type 'A', the start symbol within a slot must be zero and the length can be from 4 to 14 (for normal CP) and up to 12 (for extended CP). For PUSCH mapping type 'B', the start symbol can be from any symbol in the slot
- Slots in a frame used for the PUSCH
- Period of the allocation in slots. If this is empty it indicates no repetition
- The allocated PRBs are relative to the BWP

```

pusch(1).PUSCHMappingType = 'A'; % PUSCH mapping type ('A'(slot-wise), 'B'(non slot-wise))
pusch(1).AllocatedSymbols = 0:13; % Range of symbols in a slot
pusch(1).AllocatedSlots = [0 1]; % Allocated slots indices
pusch(1).AllocatedPeriod = 5; % Allocation period in slots (empty implies no repetition)
pusch(1).AllocatedPRB = 0:10; % PRB allocation

```

### DM-RS Configuration

The following DM-RS parameters can be set

```

% DM-RS configuration (TS 38.211 section 6.4.1.1)
pusch(1).DMRSConfigurationType = 1; % DM-RS configuration type (1,2)
pusch(1).NumCDMGroupsWithoutData = 2; % Number of DM-RS CDM groups without data. The v
pusch(1).PortSet = [0 2]; % DM-RS ports to use for the layers. The number
pusch(1).DMRSTypeAPosition = 2; % Mapping type A only. First DM-RS symbol posit
pusch(1).DMRSLength = 1; % Number of front-loaded DM-RS symbols (1(single) or 2)
pusch(1).DMRSAdditionalPosition = 2; % Additional DM-RS symbol positions (max range (0
pusch(1).NIDNSCID = 1; % Scrambling identity for CP-OFDM (0...65535). U
pusch(1).NSCID = 0; % Scrambling initialization for CP-OFDM (0,1)
pusch(1).NRSID = 0; % Scrambling identity for DFT-s-OFDM DM-RS (0...65535)

```



```
pusch(1).PowerDMRS = 0;           % Additional power boosting in dB
pusch(1).GroupHopping = 'enable'; % {'enable','disable','neither'}. This parameter
```

The parameter `GroupHopping` is used in DM-RS sequence generation when transform precoding is enabled. This can be set to

- 'enable' to indicate the presence of group hopping. It is configured by higher-layer parameter *sequenceGroupHopping*
- 'disable' to indicate the presence of sequence hopping. It is configured by higher-layer parameter *sequenceHopping*
- 'neither' to indicate both group hopping and sequence hopping are not present

Note: The number of DM-RS CDM groups without data depends on the configuration type. The maximum number of DM-RS CDM groups can be 2 for DM-RS configuration type 1 and it can be 3 for DM-RS configuration type 2.

### UCI on PUSCH

The following parameters must be set to transmit UCI on PUSCH in overlapping slots:

- Disable UL-SCH transmission on the overlapping slots of PUSCH (1/0). When set to 1, UL-SCH transmission is disabled on PUSCH. The example considers there is UL-SCH transmission all the time on PUSCH. A provision is provided to disable the UL-SCH transmission on the overlapping slots of PUSCH and PUCCH
- `BetaOffsetACK`, `BetaOffsetCSI1` and `BetaOffsetCSI2` can be set from the tables 9.3-1, 9.3-2 TS 38.213 Section 9.3
- `ScalingFactor` is provided by higher layer parameter *scaling*, as per TS 38.212, Section 6.3.2.4. The possible value is one of the set {0.5, 0.65, 0.8, 1}. This is used to limit the number of resource elements assigned to UCI on PUSCH

```
pusch(1).DisableULSCH = 1;           % Disable UL-SCH on overlapping slots of PUSCH a
pusch(1).BetaOffsetACK = 1;          % Power factor of HARQ-ACK
pusch(1).BetaOffsetCSI1 = 2;         % Power factor of CSI part 1
pusch(1).BetaOffsetCSI2 = 2;         % Power factor of CSI part 2
pusch(1).ScalingFactor = 1;          % Scaling factor (0.5, 0.65, 0.8, 1)
```

### Specifying Multiple PUSCH Instances

A second PUSCH sequence instance is specified next using the second BWP.

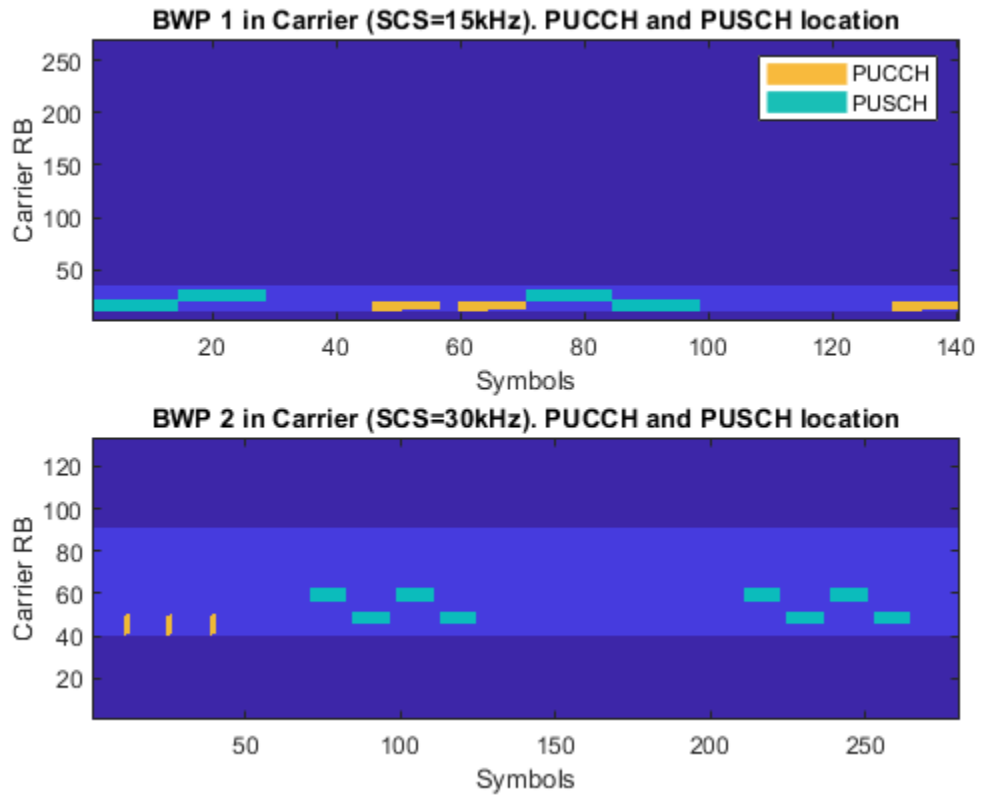
```
pusch(2) = pusch(1);
pusch(2).Enable = 1;
```

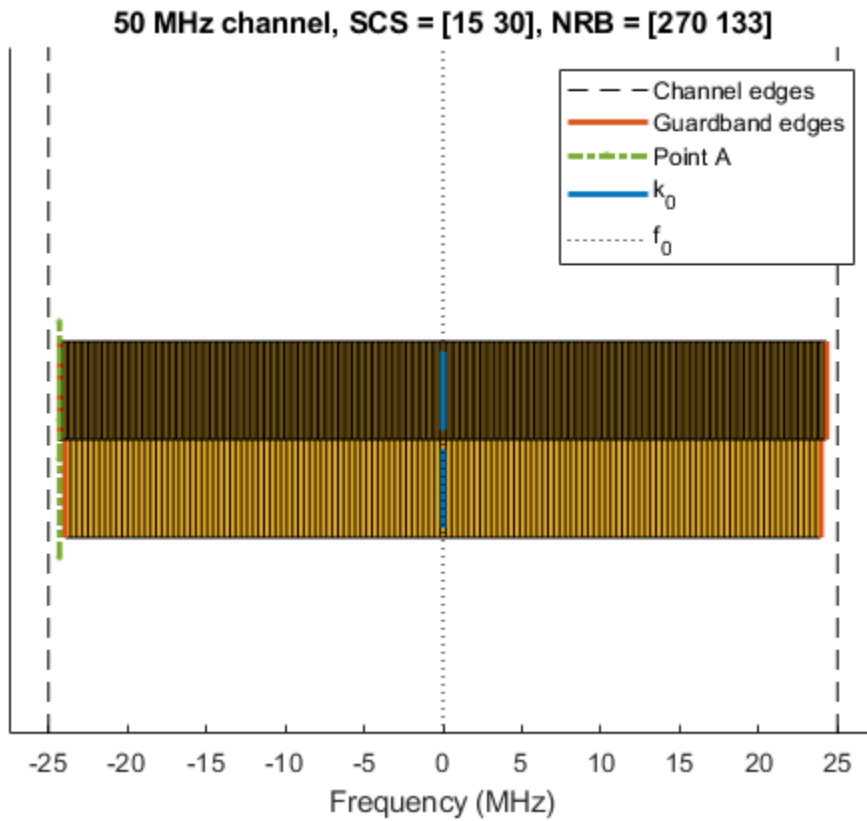
```
pusch(2).BWP = 2;  
pusch(2).AllocatedSymbols = 0:11;  
pusch(2).AllocatedSlots = [5 6 7 8];  
pusch(2).AllocatedPRB = 5:10;  
pusch(2).AllocatedPeriod = 10;  
pusch(2).TransformPrecoding = 1;  
pusch(2).IntraSlotFreqHopping = 'disabled';  
pusch(2).GroupHopping = 'neither';  
pusch(2).NLayers = 1;  
pusch(2).PortSet = 1;  
pusch(2).RNTI = 0;
```

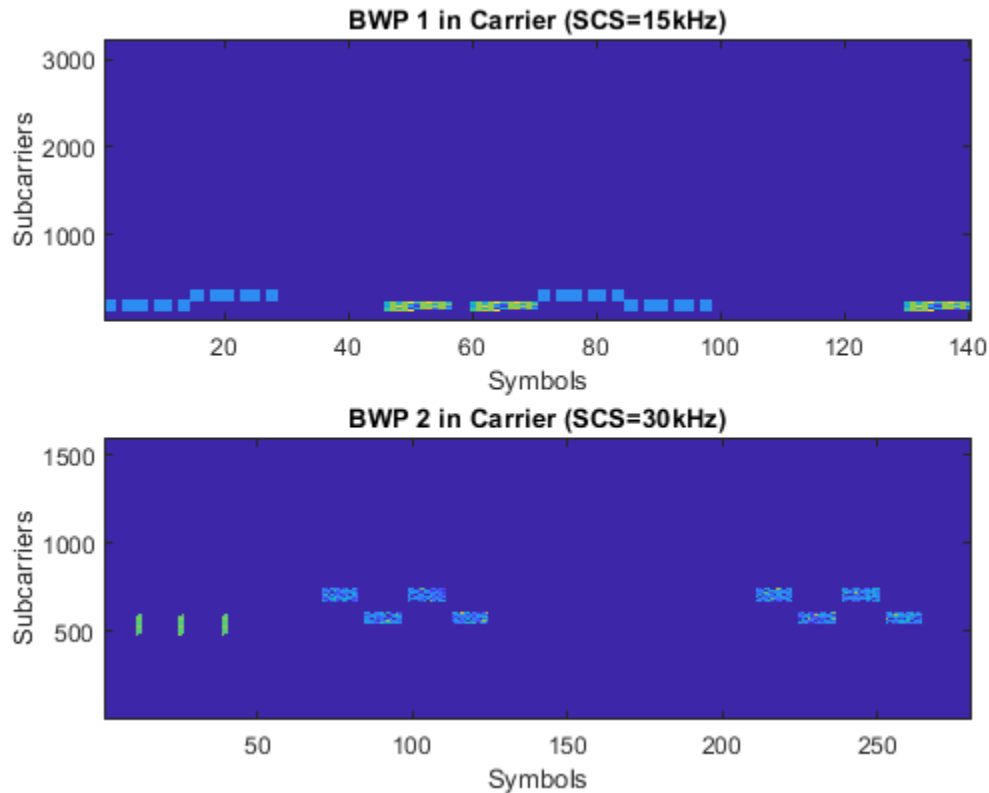
### **Waveform Generation**

This section collects all the parameters into the carrier configuration and generates the waveform.

```
% Collect together channel oriented parameter sets into a single  
% configuration  
waveconfig.Carriers = carriers;  
waveconfig.BWP = bwp;  
waveconfig.PUCCH = pucch;  
waveconfig.PUSCH = pusch;  
  
% Generate complex baseband waveform  
[waveform,bwpset] = hNRUplinkWaveformGenerator(waveconfig);
```







The waveform generator also plots the SCS carrier alignment and the resource grids for the bandwidth parts (this is controlled by the field `DisplayGrids` in the carrier configuration). The following plots are generated:

- Resource grid showing the location of the components (PUCCH and PUSCH) in each BWP. This does not plot the power of the signals, just their location in the grid
- Schematic diagram of SCS carrier alignment with the associated guardbands
- Generated waveform in the frequency domain for each BWP. This includes the PUCCH and PUSCH instances

The waveform generator function returns the time domain waveform and a struct array `bwpsset`, which contains the following fields:

- The resource grid corresponding to this BWP
- The resource grid of the overall bandwidth containing the channels and signals in this BWP
- An info structure with information corresponding to the BWP. The contents of this info structure for the first BWP are shown below:

```
disp('Information associated to BWP 1:')  
disp(bwpset(1).Info)
```

```
Information associated to BWP 1:  
    SamplingRate: 61440000  
        Nfft: 4096  
        Windowing: 10  
CyclicPrefixLengths: [1×14 double]  
    SymbolLengths: [1×14 double]  
        NSubcarriers: 3240  
    SubcarrierSpacing: 15  
    SymbolsPerSlot: 14  
    SlotsPerSubframe: 1  
    SymbolsPerSubframe: 14  
    SamplesPerSubframe: 61440  
    SubframePeriod: 1.0000e-03  
        Midpoints: [1×141 double]  
    WindowOverlap: [10 10 10 10 10 10 10 10 10 10 10 10 10 10]  
        k0: 0
```

Note that the generated resource grid is a 3D matrix where the different planes represent the antenna ports. For the different physical channels and signals the lowest port is mapped to the first plane of the grid.

## See Also

### System Objects

nrULSCH

### Functions

nrPUCCH0 | nrPUCCH1 | nrPUCCH2 | nrPUCCH3 | nrPUCCH4 | nrPUSCH

## **More About**

- “5G NR Downlink Carrier Waveform Generation” on page 3-2





### Primary Synchronization Signal (PSS)

Create the PSS for a given cell identity:

```
ncellid = 17;
pssSymbols = nrPSS(ncellid)
```

```
pssSymbols = 127×1
```

```
-1
-1
-1
-1
-1
-1
-1
1
1
1
-1
⋮
```

The variable `pssSymbols` is a column vector containing the 127 BPSK symbols of the PSS.

Create the PSS indices:

```
pssIndices = nrPSSIndices;
```

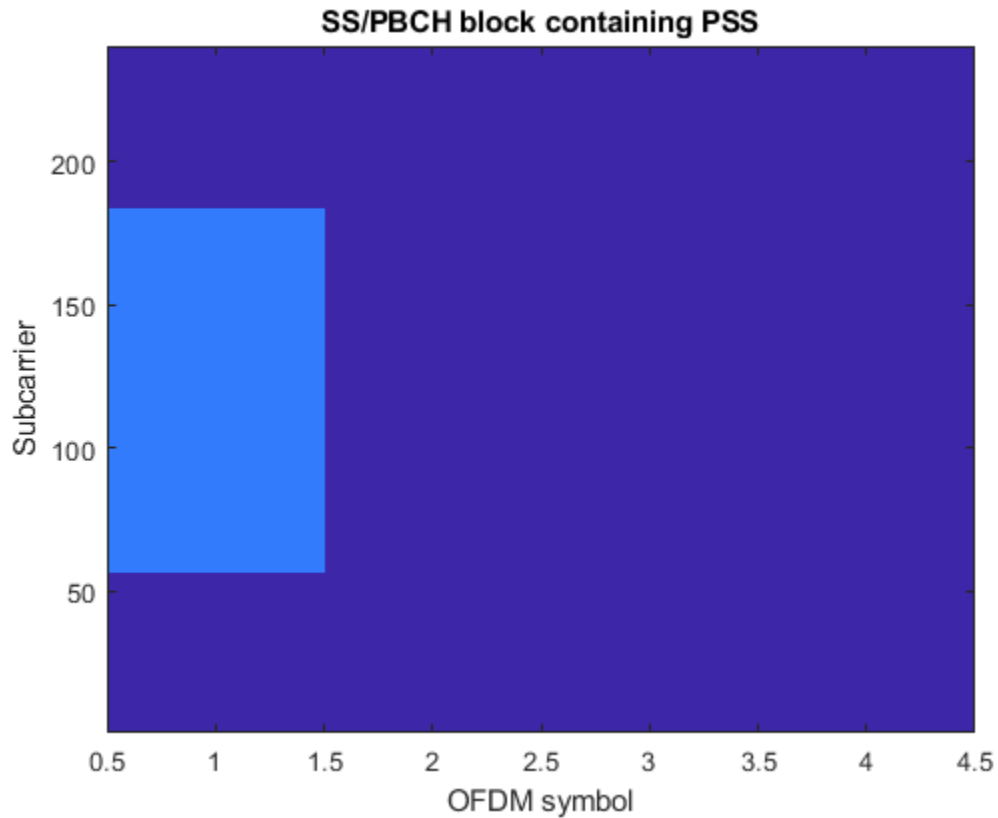
The variable `pssIndices` is a column vector of the same size as `pssSymbols`. The value in each element of `pssIndices` is the linear index of the location in the SS/PBCH block to which the corresponding symbols in `pssSymbols` should be mapped. Therefore the mapping of the PSS symbols to the SS/PBCH block can be performed with a simple MATLAB assignment, using linear indexing to select the correct elements of the SS/PBCH block matrix:

```
ssblock(pssIndices) = pssSymbols;
```

Plot the SS/PBCH block matrix to show the location of the PSS:

```
imagesc(abs(ssblock));
caxis([0 4]);
axis xy;
xlabel('OFDM symbol');
```

```
ylabel('Subcarrier');  
title('SS/PBCH block containing PSS');
```



Secondary Synchronization Signal (SSS)

Create the SSS for the same cell identity as configured for the PSS:

```
sssSymbols = nrSSS(ncellid)
```

```
sssSymbols = 127x1
```

```
-1  
1  
-1  
-1
```

```

-1
 1
-1
 1
-1
 1
  :

```

Create the SSS indices and map the SSS symbols to the SS/PBCH block, following the same pattern used for the PSS. Note that a scaling factor of 2 is applied to the SSS symbols, to represent  $\beta_{\text{SSS}}$  in TS 38.211 Section 7.4.3.1.2:

```

sssIndices = nrSSSIndices;
ssblock(sssIndices) = 2 * sssSymbols;

```

The default form of the indices is 1-based linear indices, suitable for linear indexing of MATLAB matrices like `ssblock` as already shown. However, the NR standard documents describe the OFDM resources in terms of OFDM subcarrier and symbol subscripts, using 0-based numbering. For convenient cross-checking with the NR standard, the indices functions accept options to allow the indexing style (linear index versus subscript) and base (0-based versus 1-based) to be selected:

```

sssSubscripts = nrSSSIndices('IndexStyle', 'subscript', 'IndexBase', '0based')

```

```

sssSubscripts = 127x3 uint32 matrix

```

```

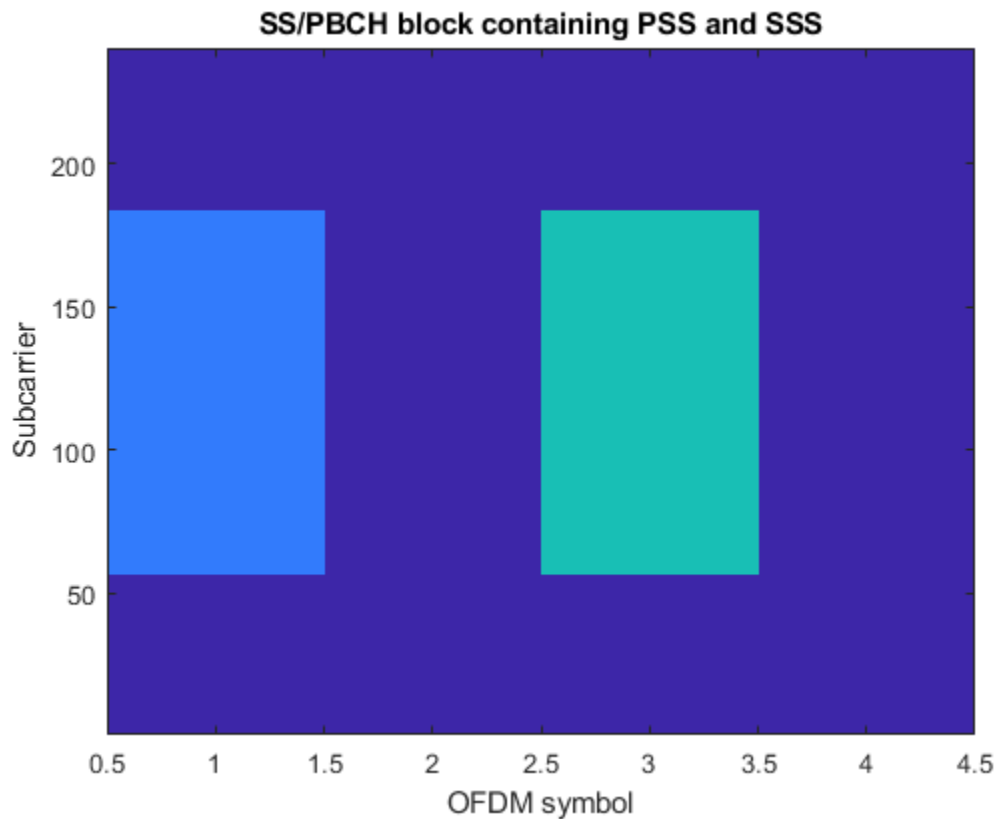
56    2    0
57    2    0
58    2    0
59    2    0
60    2    0
61    2    0
62    2    0
63    2    0
64    2    0
65    2    0
  :

```

It can be seen from the subscripts that the SSS is located in OFDM symbol 2 (0-based) of the SS/PBCH block, starting at subcarrier 56 (0-based).

Plot the SS/PBCH block matrix again to show the locations of the PSS and SSS:

```
imagesc(abs(ssblock));  
caxis([0 4]);  
axis xy;  
xlabel('OFDM symbol');  
ylabel('Subcarrier');  
title('SS/PBCH block containing PSS and SSS');
```



### Physical Broadcast Channel (PBCH)

The PBCH carries a codeword of length 864 bits, created by performing BCH encoding of the master information block (MIB). For more information on BCH coding, see the functions `nrBCH` and `nrBCHDecode` and their use in the “NR Synchronization Procedures” example. Here a PBCH codeword consisting of 864 random bits is used:

```
cw = randi([0 1],864,1);
```

The PBCH modulation consists of the following steps as described in TS 38.211 Section 7.3.3:

- Scrambling
- Modulation
- Mapping to physical resources

#### Scrambling and modulation

Multiple SS/PBCH blocks are transmitted across half a frame, as described in the cell search procedure in TS 38.213 Section 4.1. Each SS/PBCH block is given an index from  $0 \dots L - 1$ , where  $L$  is the number SS/PBCH blocks in the half frame. The scrambling sequence for the PBCH is initialized according to the cell identity `ncellid`, and the subsequence used to scramble the PBCH codeword depends on the value  $v$ , 2 or 3 LSBs of SS/PBCH block index, as described in TS 38.211 Section 7.3.3.1. In this example,  $v = 0$  is used. The function `nrPBCH` creates the appropriate subsequence of the scrambling sequence, performs scrambling and then performs QPSK modulation:

```
v = 0;
pbchSymbols = nrPBCH(cw,ncellid,v)
```

```
pbchSymbols = 432x1 complex
```

```
-0.7071 + 0.7071i
-0.7071 + 0.7071i
-0.7071 + 0.7071i
-0.7071 - 0.7071i
 0.7071 + 0.7071i
-0.7071 + 0.7071i
-0.7071 + 0.7071i
 0.7071 - 0.7071i
 0.7071 + 0.7071i
 0.7071 + 0.7071i
  :
```

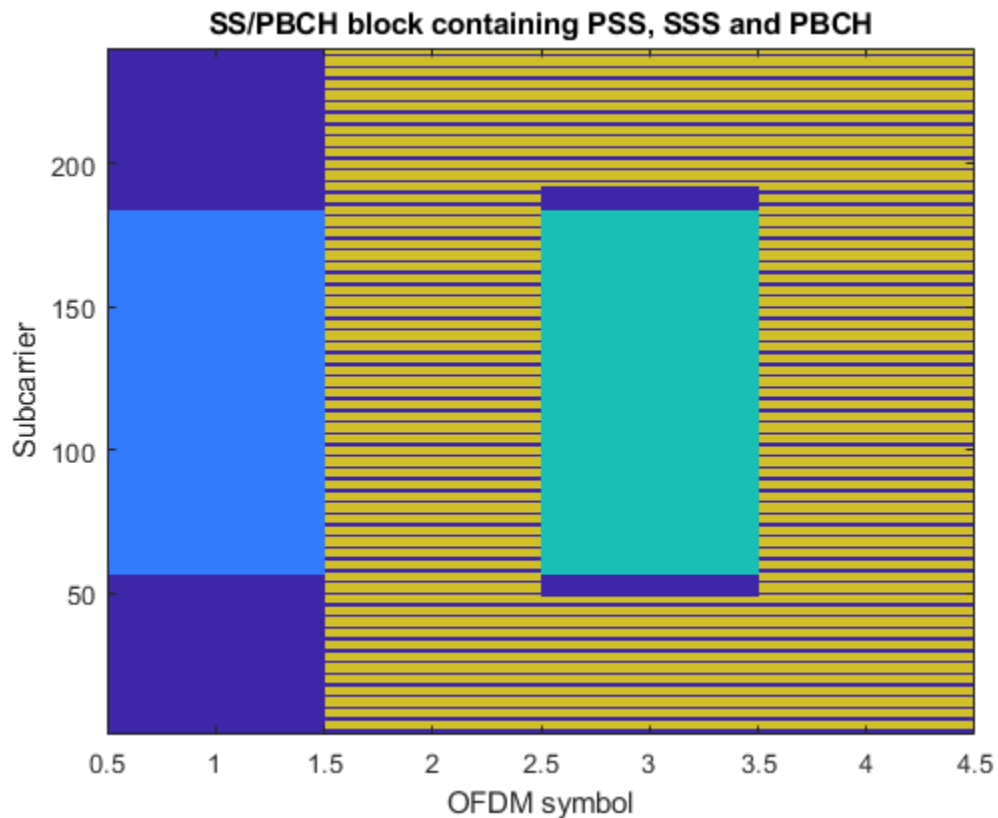
#### Mapping to resource elements

Create the PBCH indices and map the PBCH symbols to the SS/PBCH block:

```
pbchIndices = nrPBCHIndices(ncellid);
ssblock(pbchIndices) = 3 * pbchSymbols;
```

Plot the SS/PBCH block matrix again to show the locations of the PSS, SSS and PBCH:

```
imagesc(abs(ssblock));
caxis([0 4]);
axis xy;
xlabel('OFDM symbol');
ylabel('Subcarrier');
title('SS/PBCH block containing PSS, SSS and PBCH');
```



PBCH Demodulation Reference Signal (PBCH DM-RS)

The final component of the SS/PBCH block is the DM-RS associated with the PBCH. Similar to the PBCH, the DM-RS sequence used derives from the SS/PBCH block index

and is configured using the variable  $\bar{i}_{\text{SSB}}$  described in TS 38.211 Section 7.4.1.4.1. Here  $\bar{i}_{\text{SSB}} = 0$  is used:

```

ibar_SSB = 0;
dmrsSymbols = nrPBCHDMRS(ncellid,ibar_SSB)

dmrsSymbols = 144x1 complex

    0.7071 - 0.7071i
    0.7071 + 0.7071i
   -0.7071 + 0.7071i
   -0.7071 + 0.7071i
    0.7071 - 0.7071i
    0.7071 + 0.7071i
    0.7071 - 0.7071i
   -0.7071 - 0.7071i
   -0.7071 - 0.7071i
    0.7071 + 0.7071i
    :

```

Note that TS 38.211 Section 7.4.1.4.1 defines an intermediate variable  $i_{\text{SSB}}$  which is defined identically to  $v$  described previously for the PBCH.

Create the PBCH DM-RS indices and map the PBCH DM-RS symbols to the SS/PBCH block:

```

dmrsIndices = nrPBCHDMRSIndices(ncellid);
ssblock(dmrsIndices) = 4 * dmrsSymbols;

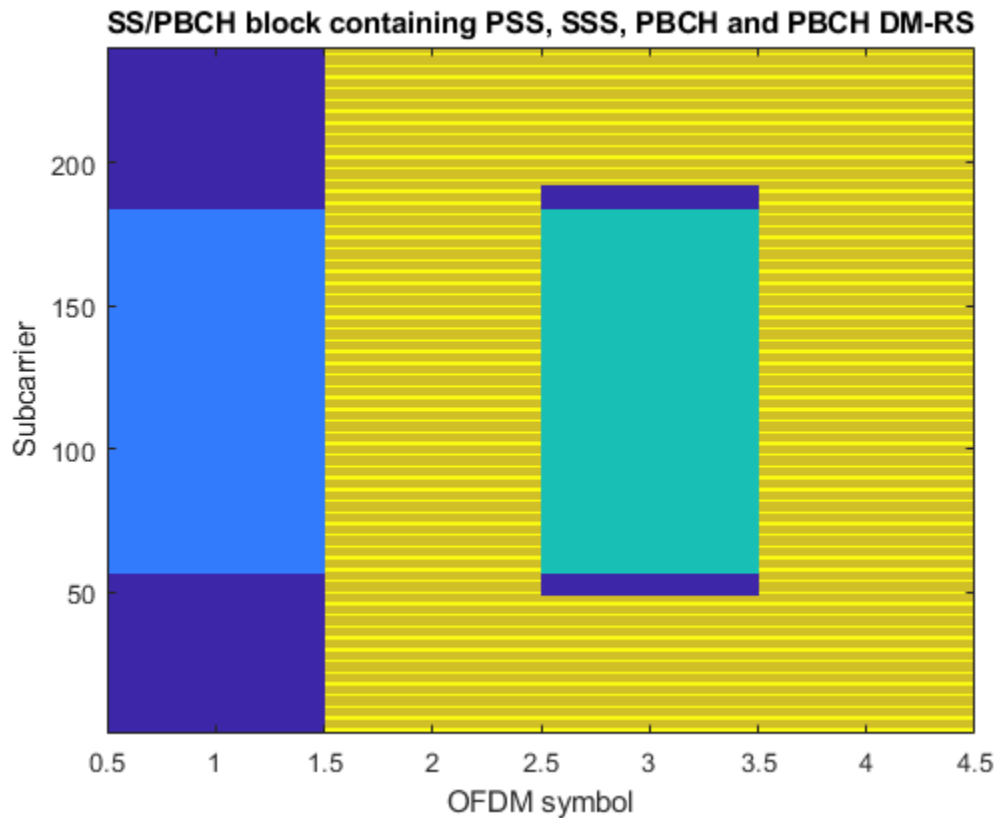
```

Plot the SS/PBCH block matrix again to show the locations of the PSS, SSS, PBCH and PBCH DM-RS:

```

imagesc(abs(ssblock));
caxis([0 4]);
axis xy;
xlabel('OFDM symbol');
ylabel('Subcarrier');
title('SS/PBCH block containing PSS, SSS, PBCH and PBCH DM-RS');

```



### Generating an SS burst

An SS burst, consisting of multiple SS/PBCH blocks, can be generated by creating a larger grid and mapping SS/PBCH blocks into the appropriate locations, with each SS/PBCH block having the correct parameters according to the location.

Create SS burst grid

In the NR standard, OFDM symbols are grouped into slots, subframes and frames. As defined in TS 38.211 Section 4.3.1, there are 10 subframes in a frame, and each subframe has a fixed duration of 1ms. Each SS burst has a duration of half a frame, and therefore spans 5 subframes:

$$n_{\text{Subframes}} = 5$$



```
nSubframes = 5
```

TS 38.211 Section 4.3.2 defines each slot as having 14 OFDM symbols (for normal cyclic prefix length) and this is fixed:

```
symbolsPerSlot = 14
```

```
symbolsPerSlot = 14
```

However, the number of slots per subframe varies and is a function of the subcarrier spacing. As the subcarrier spacing increases, the OFDM symbol duration decreases and therefore more OFDM symbols can be fitted into the fixed subframe duration of 1ms.

There are 5 subcarrier spacing configurations  $\mu = 0 \dots 4$ , with the corresponding subcarrier spacing being  $15 \cdot 2^\mu$  kHz. In this example we shall use  $\mu = 1$ , corresponding to 30 kHz subcarrier spacing:

```
mu = 1
```

```
mu = 1
```

The number of slots per subframe is  $2^\mu$ , as doubling the subcarrier spacing halves the OFDM symbol duration. Note that definition of a slot in NR is different from LTE: a subframe in LTE consists of 2 slots of 7 symbols (for normal cyclic prefix) whereas in NR, a subframe using the LTE subcarrier spacing ( $\mu = 0$ , 15 kHz) consists of 1 slot of 14 symbols.

Calculate the total number of OFDM symbols in an SS burst:

```
nSymbols = symbolsPerSlot * 2^mu * nSubframes
```

```
nSymbols = 140
```

Create an empty grid for the whole SS burst :

```
ssburst = zeros([240 nSymbols])
```

```
ssburst = 240x140
```

```

0 0 0 0 0 0 0 0 0 0 0 0 0 0
0 0 0 0 0 0 0 0 0 0 0 0 0 0
0 0 0 0 0 0 0 0 0 0 0 0 0 0
0 0 0 0 0 0 0 0 0 0 0 0 0 0
0 0 0 0 0 0 0 0 0 0 0 0 0 0

```

```

0      0      0      0      0      0      0      0      0      0      0      0      0      0
0      0      0      0      0      0      0      0      0      0      0      0      0      0
0      0      0      0      0      0      0      0      0      0      0      0      0      0
0      0      0      0      0      0      0      0      0      0      0      0      0      0
0      0      0      0      0      0      0      0      0      0      0      0      0      0
:

```

### Define SS block pattern

The pattern of SS/PBCH blocks within an SS burst is indirectly specified by the cell search procedure in TS 38.213, which describes the locations in which the UE may detect an SS/PBCH block. There are 5 block patterns, Case A - Case E, which have different subcarrier spacings and are applicable for different carrier frequencies.

Create the indices of the first symbols in the candidate SS/PBCH blocks for block pattern Case B, which has  $L = 8$  blocks per burst:

```

n = [0, 1];
firstSymbolIndex = [4; 8; 16; 20] + 28*n;
firstSymbolIndex = firstSymbolIndex(:).';

firstSymbolIndex = 1x8
     4     8    16    20    32    36    44    48

```

### Create SS burst content

Now a loop can be created which generates each SS block and assigns it into the appropriate location of the SS burst. Note the following:

- The code re-uses various variables created earlier in this example (PSS, SSS, and 4 sets of indices)
- The PSS and SSS are independent of the SS/PBCH block index, so can be mapped into the SS block before the loop
- The PBCH indices and PBCH DM-RS indices are independent of the SS/PBCH block index, so do not need updated in the loop
- $i_{SSB}$ ,  $\bar{i}_{SSB}$  and  $\nu$  are set up according to the rules in TS 38.211 Sections 7.3.3.1 and 7.4.1.4.1 for the case of  $L = 8$ .
- Each channel / signal has been scaled in order to give them different colors in the final plot

```

ssblock = zeros([240 4]);
ssblock(pssIndices) = pssSymbols;
ssblock(sssIndices) = 2 * sssSymbols;

for ssbIndex = 1:length(firstSymbolIndex)

    i_SSB = mod(ssbIndex,8);
    ibar_SSB = i_SSB;
    v = i_SSB;

    pbchSymbols = nrPBCH(cw,ncellid,v);
    ssblock(pbchIndices) = 3 * pbchSymbols;

    dmrsSymbols = nrPBCHDMRS(ncellid,ibar_SSB);
    ssblock(dmrsIndices) = 4 * dmrsSymbols;

    ssburst(:,firstSymbolIndex(ssbIndex) + (0:3)) = ssblock;

end

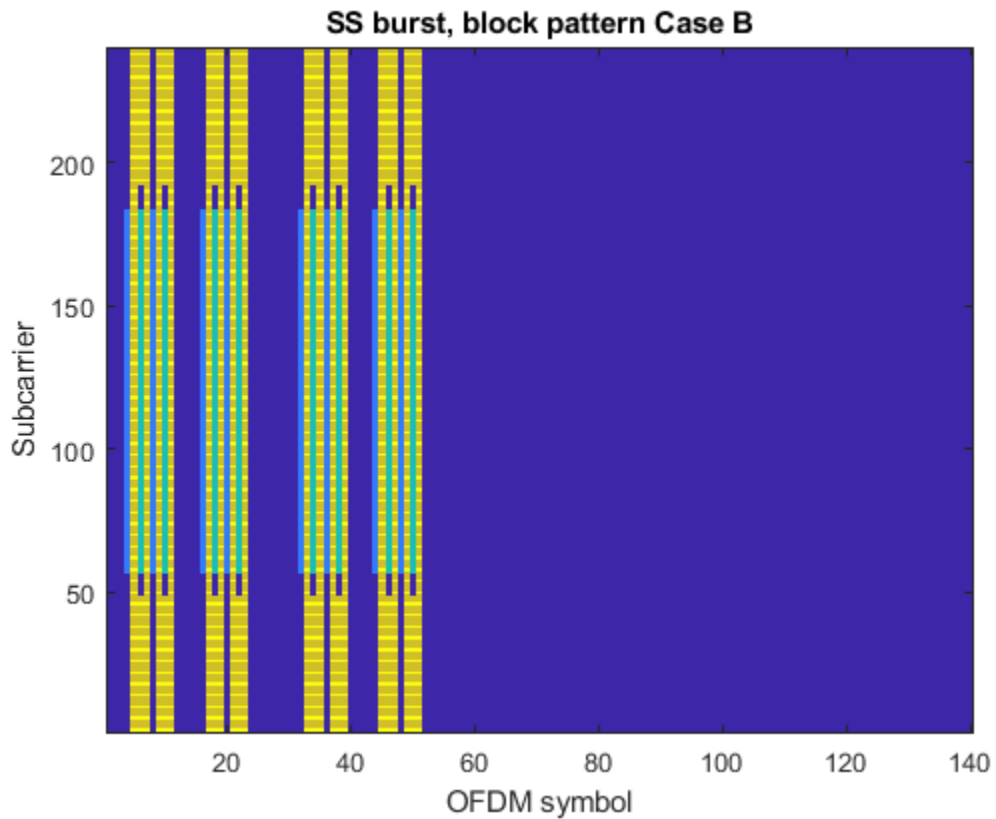
```

Finally, plot the SS burst content:

```

imagesc(abs(ssburst));
caxis([0 4]);
axis xy;
xlabel('OFDM symbol');
ylabel('Subcarrier');
title('SS burst, block pattern Case B');

```



## See Also

### Functions

[nrPBCH](#) | [nrPBCHDMRS](#) | [nrPBCHDMRSIndices](#) | [nrPBCHIndices](#) | [nrPSS](#) | [nrPSSIndices](#) | [nrSSS](#) | [nrSSSIndices](#)

### More About

- “NR Synchronization Procedures”

## Modeling Downlink Control Information

This example describes the downlink control information (DCI) processing for the 5G New Radio communications system. Starting from a random DCI message, it models the message encoding followed by the physical downlink control channel (PDCCH) processing on the transmit end. Corresponding receiver components recover the transmitted control information elements.

### System Parameters

Set parameters for a UE-specific search space.

```
rng(211);           % Set RNG state for repeatability

nID = 23;           % pdcch-DMRS-ScramblingID
rnti = 100;         % C-RNTI for PDCCH in a UE-specific search space
K = 64;             % Number of DCI message bits
E = 288;            % Number of bits for PDCCH resources
```

### DCI Encoding

The DCI message bits based on a downlink format are encoded using the `nrDCIEncode` function, which includes the stages of CRC attachment, polar encoding and rate matching.

```
dciBits = randi([0 1],K,1,'int8');
dciCW = nrDCIEncode(dciBits,rnti,E);
```

### PDCCH Symbol Generation

The encoded DCI bits (a codeword) are mapped onto the physical downlink control channel (PDCCH) using the `nrPDCCH` function which generates the scrambled, QPSK-modulated symbols. The scrambling accounts for the user-specific parameters.

```
sym = nrPDCCH(dciCW,nID,rnti);
```

For NR, the PDCCH symbols are then mapped to the resource elements of an OFDM grid which also has PDSCH, PBCH and other reference signal elements. These are followed by OFDM modulation and transmission over a channel. For simplicity, we directly pass the PDCCH symbols over an AWGN channel next.

### Channel

The PDCCH symbols are transmitted over an AWGN channel with a specified SNR, accounting for the coding rate and QPSK modulation.

```
EbNo = 3; % in dB
bps = 2; % bits per symbol, 2 for QPSK
EsNo = EbNo + 10*log10(bps);
snrdb = EsNo + 10*log10(K/E);

rxSym = awgn(sym,snrdb,'measured');
```

### PDCCH Decoding

The received symbols are demodulated with known user-specific parameters and channel noise variance using the `nrPDCCHDecode` function. The soft output is the log-likelihood ratio for each bit in the codeword.

```
noiseVar = 10.^(-snrdb/10); % assumes unit signal power
rxCW = nrPDCCHDecode(rxSym,nID,rnti,noiseVar);
```

### DCI Decoding

An instance of the received PDCCH codeword is then decoded by the `nrDCIDecode` function. This includes the stages of rate recovery, polar decoding and CRC decoding to recover the transmitted information bits.

```
listLen = 8; % polar decoding list length
[decDCIBits,mask] = nrDCIDecode(rxCW,K,listLen);

isequal(mask,rnti)
isequal(decDCIBits,dciBits)
```

```
ans =
```

```
logical
```

```
1
```

```
ans =
```

```
logical
```

1

The output mask matches the C-RNTI used for the UE, thereby confirming the message recipient. For the chosen system parameters, the decoded information matches the transmitted information bits.

## See Also

### Functions

`nrDCIDecode` | `nrDCIEncode` | `nrPDCCH` | `nrPDCCHDecode`

### More About

- “Downlink Control Processing and Procedures”

## 5G New Radio Polar Coding

This example highlights the new polar channel coding technique chosen for 5G New Radio (NR) communications system. Of the two main types of code constructions specified by 3GPP, this example models the CRC-Aided Polar (CA-Polar) coding scheme. This example describes the main components of the polar coding scheme with individual components for code construction, encoding and decoding along-with rate-matching. It models a polar-coded QPSK-modulated link over AWGN and presents Block-Error-Rate results for different message lengths and code rates for the coding scheme.

### Introduction

The selection of polar codes as the channel coding technique for control channels for 5G NR communications system has proven the merits of Arikan's [ 1 ] discovery and will establish their application in commercial systems [ 6 ]. Based on the concept of channel polarization, this new coding family is capacity achieving as opposed to just capacity approaching. With better or comparable performance than LDPC and turbo codes, it supersedes the tail-biting convolutional codes used in LTE systems for control channels. It is applied for downlink and uplink control information (DCI/UCI) for the enhanced mobile broadband (eMBB) use case, as well as the broadcast channel (BCH). Alternatively, the channel coding scheme for data channels for eMBB is specified to be flexible LDPC for all block sizes.

This example highlights the components to enable a polar coding downlink simulation using QPSK modulation over an AWGN channel. In the following sections, the individual polar coding components are further detailed.

```
s = rng(100);           % Seed the RNG for repeatability
```

Specify the code parameters used for a simulation.

```
% Code parameters
```

```
K = 54;                % Message length in bits, including CRC, K > 30
```

```
E = 124;              % Rate matched output length, E <= 8192
```

```
EbNo = 0.8;           % EbNo in dB
```

```
L = 8;                % List length, a power of two, [1 2 4 8]
```

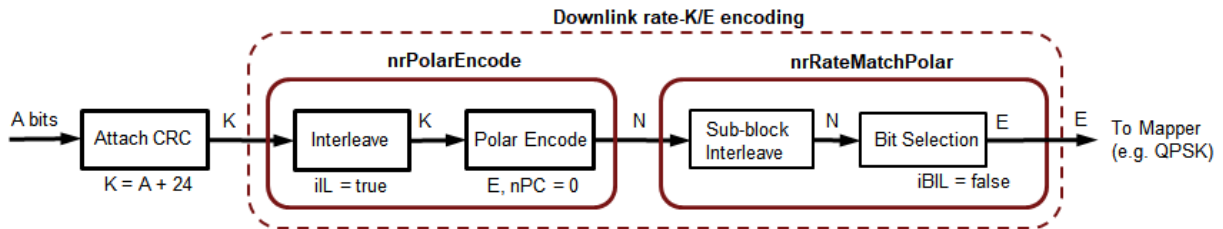
```
numFrames = 10;       % Number of frames to simulate
```

```
linkDir = 'DL';       % Link direction: downlink ('DL') OR uplink ('UL')
```



## Polar Encoding

The following schematic details the transmit-end processing for the downlink, with relevant components and their parameters highlighted.



For the downlink, the input bits are interleaved prior to polar encoding. The CRC bits appended at the end of the information bits are thus distributed for the CA-Polar scheme. This interleaving is not specified for the uplink.

The polar encoding uses an SNR-independent method where the reliability of each subchannel is computed offline and the ordered sequence stored for a maximum code length [ 6 ]. The nested property of polar codes allows this sequence to be used for any code rate and all code lengths smaller than the maximum code length.

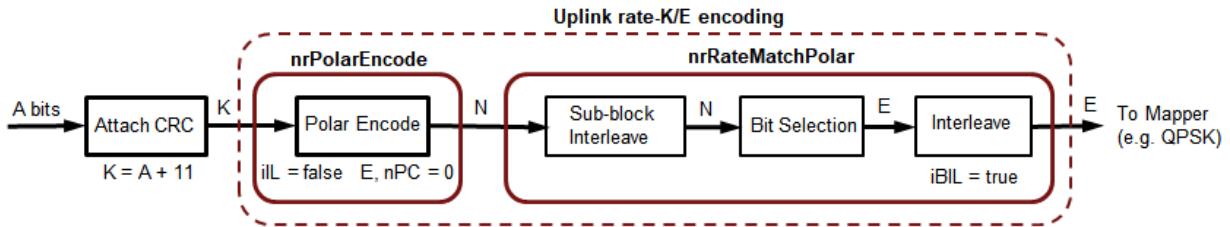
This sequence is computed for given rate-matched output length,  $E$ , and information length,  $K$ , by the function `nrPolarEncode`, which implements the non-systematic encoding of the input  $K$  bits.

```

if strcmp(linkDir, 'DL')
    % Downlink scenario (K >= 36, including CRC bits)
    crcLen = 24;           % Number of CRC bits for DL, Section 5.1, [6]
    poly = '24C';        % CRC polynomial
    nPC = 0;             % Number of parity check bits, Section 5.3.1.2, [6]
    nMax = 9;           % Maximum value of n, for 2^n, Section 7.3.3, [6]
    iIL = true;         % Interleave input, Section 5.3.1.1, [6]
    iBIL = false;      % Interleave coded bits, Section 5.4.1.3, [6]
else
    % Uplink scenario (K > 30, including CRC bits)
    crcLen = 11;
    poly = '11';
    nPC = 0;
    nMax = 10;
    iIL = false;
    iBIL = true;
end

```

The following schematic details the transmit-end processing for the uplink, for a payload size greater than 19 bits and no code-block segmentation, with relevant components and their parameters highlighted.



### Rate Matching and Rate Recovery

The polar encoded set of bits ( $N$ ) are rate-matched to output the specified number of bits ( $E$ ) for resource element mapping [ 7 ]. The coded bits are sub-block interleaved and passed to a circular buffer of length  $N$ . Depending on the desired code rate and selected values of  $K$ ,  $E$ , and  $N$ , either of repetition ( $E \geq N$ ), and puncturing or shortening ( $E < N$ ) is realized by reading the output bits from the buffer.

- For puncturing,  $E$  bits are taken from the end
- For shortening,  $E$  bits are taken from the start
- For repetition,  $E$  bits are repeated modulo  $N$ .

For the downlink, the selected bits are passed on to the modulation mapper, while for the uplink, they are further interleaved prior to mapping. The rate-matching processing is implemented by the function `nrRateMatchPolar`.

At the receiver end, rate recovery is accomplished for each of the cases

- For puncturing, corresponding LLRs for the bits removed are set to zero
- For shortening, corresponding LLRs for the bits removed are set to a large value
- For repetition, the set of LLRs corresponding to first  $N$  bits are selected.

The rate-recovery processing is implemented by the function `nrRateRecoverPolar`.

```
R = K/E; % Effective code rate
bps = 2; % bits per symbol, 1 for BPSK, 2 for QPSK
EsNo = EbNo + 10*log10(bps);
snrdB = EsNo + 10*log10(R); % in dB
noiseVar = 1./(10.^(snrdB/10));
```

```
% Channel
chan = comm.AWGNChannel('NoiseMethod','Variance','Variance',noiseVar);
```

### Polar Decoding

The implicit CRC encoding of the downlink (DCI or BCH) or uplink (UCI) message bits dictates the use of the CRC-Aided Successive Cancellation List Decoding (CA-SCL) [ 3 ] as the channel decoder algorithm. It is well known that CA-SCL decoding can outperform turbo or LDPC codes and this was one of the major factors in the adoption of polar codes by 3GPP.

Tal & Vardy [ 2 ] describe the SCL decoding algorithm in terms of likelihoods (probabilities). However, due to underflow, the inherent computations are numerically unstable. To overcome this issue, Stimming et.al. [ 5 ] offer the SCL decoding solely in the log-likelihood ratio (LLR) domain. The list decoding is characterized by the L parameter, which represents the number of most likely decoding paths retained. At the end of the decoding, the most likely code-path among the L paths is the decoder output. As L is increased, the decoder performance also improves, however, with a diminishing-returns effect.

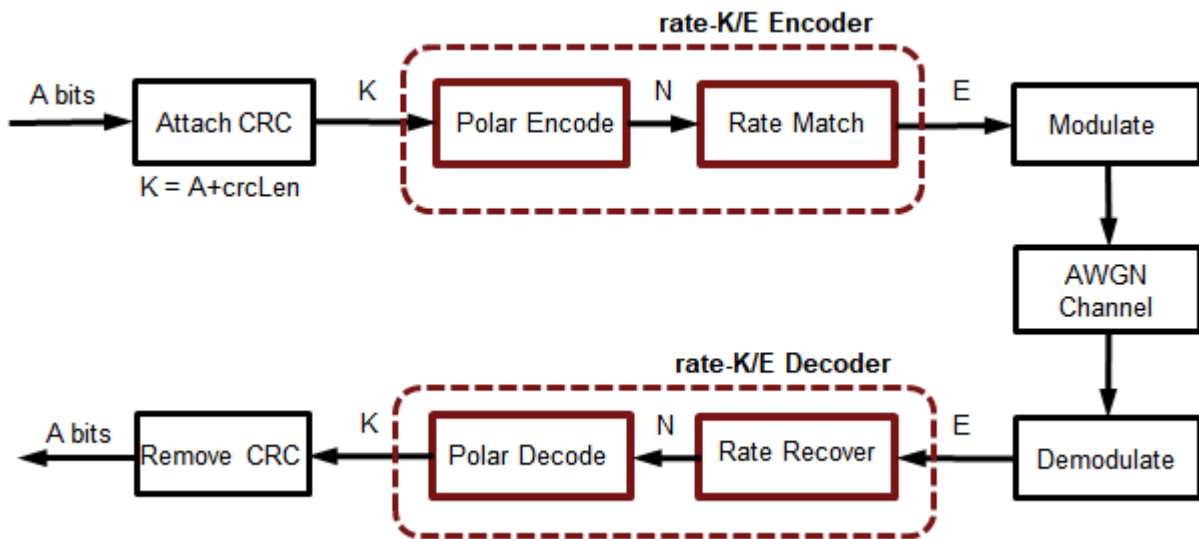
For an input message which is concatenated with a CRC, CA-SCL decoding prunes out any of the paths for which the CRC is invalid, if at least one path has the correct CRC. This additional insight in the final path selection improves the performance further, when compared to SCL decoding. For the downlink, a CRC of 24 bits is used, while for the uplink CRCs of 6 and 11 bits are specified, which vary on the value of K.

The decoder is implemented by the function `nrPolarDecode`, which supports all three CRC lengths. The decoder function also accounts for the input bit interleaving specified at the transmitter for the downlink, prior to outputting the decoded bits.

```
% Error meter
ber = comm.ErrorRate;
```

### Frame Processing Loop

This section shows how the prior described components for polar coding are used in a Block Error Rate (BLER) simulation. The simulation link is highlighted in the following schematic.



For each frame processed, the following steps are performed:

- $K - \text{crcLen}$  random bits are generated,
- A CRC is computed and appended to these bits
- The CRC appended bits are polar encoded to the mother code block length
- Rate-matching is performed to transmit  $E$  bits
- The  $E$  bits are QPSK modulated
- White Gaussian Noise of specified power is added
- The noisy signal is soft QPSK demodulated to output LLR values
- Rate recovery is performed accounting for either of puncturing, shortening or repetition
- The recovered LLR values are polar decoded using the CA-SCL algorithm, including deinterleaving.
- Off the decoded  $K$  bits, the first  $K - \text{crcLen}$  bits are compared with those transmitted to update the BLER and bit-error-rate (BER) metrics.

At the end of the simulation, the two performance indicators, BLER and BER, are reported.

```
numferr = 0;
for i = 1:numFrames
```

```

% Generate a random message
msg = randi([0 1],K-crcLen,1);

% Attach CRC
msgcrc = nrCRCEncode(msg,poly);

% Polar encode
encOut = nrPolarEncode(msgcrc,E,nMax,iIL);
N = length(encOut);

% Rate match
modIn = nrRateMatchPolar(encOut,K,E,iBIL);

% Modulate
modOut = nrSymbolModulate(modIn,'QPSK');

% Add White Gaussian noise
rSig = chan(modOut);

% Soft demodulate
rxLLR = nrSymbolDemodulate(rSig,'QPSK',noiseVar);

% Rate recover
decIn = nrRateRecoverPolar(rxLLR,K,N,iBIL);

% Polar decode
decBits = nrPolarDecode(decIn,K,E,L,nMax,iIL,crcLen);

% Compare msg and decoded bits
errStats = ber(double(decBits(1:K-crcLen)), msg);
numferr = numferr + any(decBits(1:K-crcLen)~=msg);

end

disp(['Block Error Rate: ' num2str(numferr/numFrames) ...
      ', Bit Error Rate: ' num2str(errStats(1)) ...
      ', at SNR = ' num2str(snrdB) ' dB'])

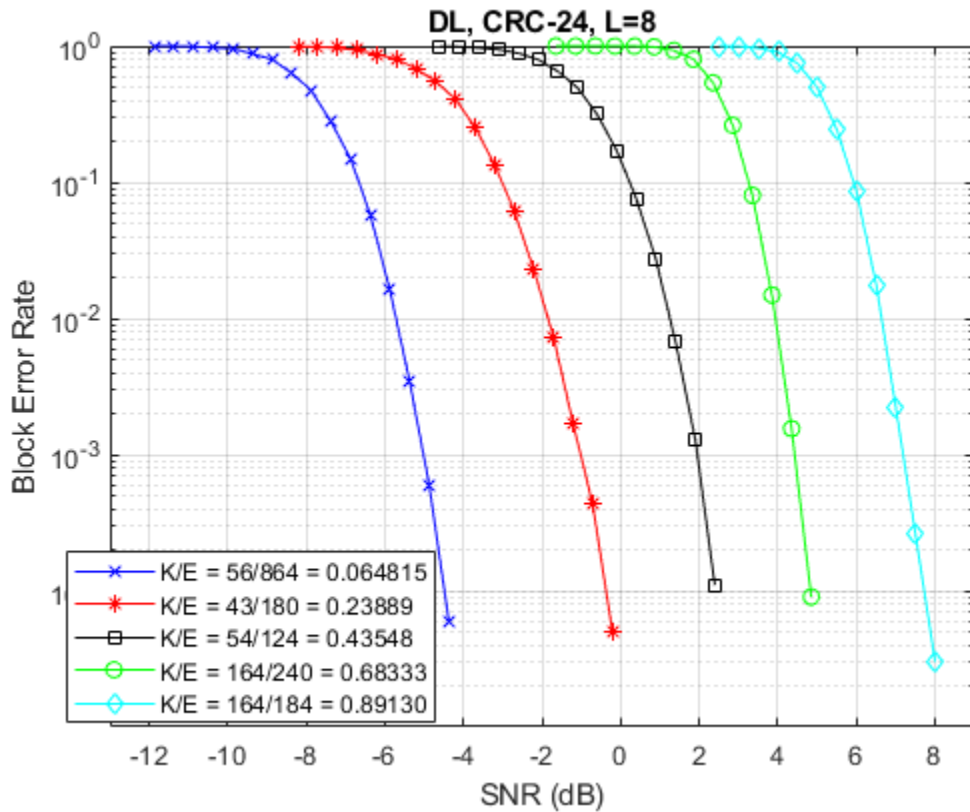
rng(s); % Restore RNG

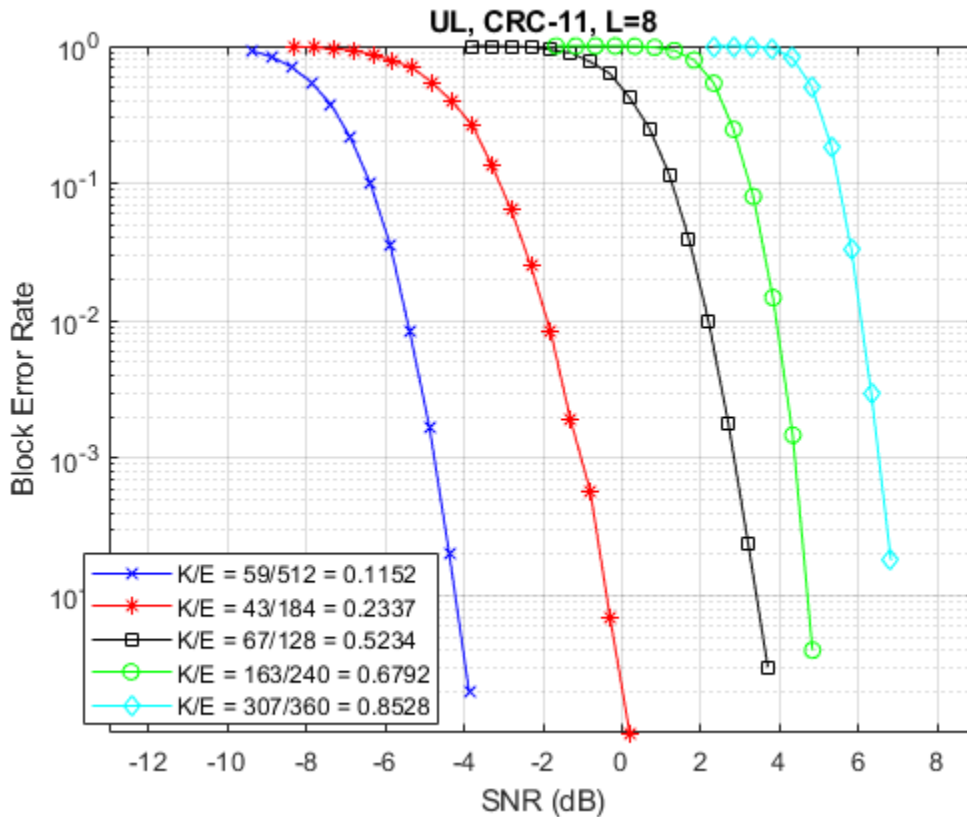
Block Error Rate: 0, Bit Error Rate: 0, at SNR = 0.20002 dB

```

## Results

To get meaningful results, simulations have to be run for a longer duration. Using scripts which encapsulate the above processing into a function that supports C-code generation, the following results for different code rates and message lengths are presented for both link directions with QPSK modulation.





The above results were generated by simulating, for each SNR point, up to 1000 frame errors or a maximum of 100e3 frames, whichever occurred first.

The BLER performance results indicate the suitability of polar codes in a communication link and their implicit support for rate-compatibility at the bit-level granularity.

The use of C-code generation tools for the components reduces the execution time, a key concern for simulations. The C-code generation is enabled by MATLAB Coder™.

### Summary and Further Exploration

This example highlights one of the polar coding schemes (CRC-Aided Polar) specified by 3GPP for New Radio control channel information (DCI, UCI) and broadcast channel (BCH). It shows the use of components for all stages of the processing (encoding, rate-

matching, rate-recovery and decoding) and uses them in a link with QPSK over an AWGN channel. Highlighted performance results for different code rates and message lengths show agreement to published trends, within parametric and simulation assumption variations.

Explore simple parameter variations (K, E, L) and their effect on BLER performance. The polar coding functions are implemented as open MATLAB® code to enable their application for both downlink/uplink control information and broadcast channel. The CA-Polar scheme is applicable for both

- Downlink, for all message lengths, and
- Uplink, for  $K > 30$ , with `crcLen = 11`, `nPC = 0`, `nMax = 10`, `iIL = false`, and `iBIL = true`.

Refer to “Modeling Downlink Control Information” on page 3-45 and “NR Synchronization Procedures” examples, for the use of polar coding functions within the DCI and BCH functions respectively.

The highlighted polar coding functions also support the Parity-Check polar coding construction and encoding. This is applicable for the uplink with UCI payloads in range  $18 \leq K \leq 25$ . This is supported by the uplink control coding functions `nRUCIEncode` and `nRUCIDecode`, which include code-block segmentation as well for appropriate values of K and E.

### **Selected References**

- 1** Arikan, E., "Channel Polarization: A Method for constructing Capacity-Achieving Codes for Symmetric Binary-Input Memoryless Channels," IEEE Transactions on Information Theory, vol. 55, No. 7, pp. 3051-3073, July 2009.
- 2** Tal, I, and Vardy, A., "List decoding of Polar Codes", IEEE Transactions on Information Theory, vol. 61, No. 5, pp. 2213-2226, May 2015.
- 3** Niu, K., and Chen, K., "CRC-Aided Decoding of Polar Codes," IEEE Communications Letters, vol. 16, No. 10, pp. 1668-1671, Oct. 2012.
- 4** Niu, K., Chen, K., and Lin, J.R., "Beyond turbo codes: rate compatible punctured polar codes", IEEE International Conference on Communications, pp. 3423-3427, 2013.
- 5** Stimming, A. B., Parizi, M. B., and Burg, A., "LLR-Based Successive Cancellation List Decoding of Polar Codes", IEEE Transaction on Signal Processing, vol. 63, No. 19, pp.5165-5179, 2015.
- 6** 3GPP TS 38.212. "NR; Multiplexing and channel coding (Release 15)." 3rd Generation Partnership Project; Technical Specification Group Radio Access Network.



- 7 R1-1711729. "WF on circular buffer of Polar Code", 3GPP TSG RAN WG1 meeting NR Ad-Hoc#2, Ericsson, Qualcomm, MediaTek, LGE. June 2017.

## See Also

### Functions

`nrPolarDecode` | `nrPolarEncode` | `nrRateMatchPolar` | `nrRateRecoverPolar`

### More About

- "Modeling Downlink Control Information" on page 3-45
- "NR Synchronization Procedures"

## LDPC Processing for DL-SCH and UL-SCH

This example highlights the low-density parity-check (LDPC) coding chain for the 5G NR downlink and uplink shared transport channels (DL-SCH and UL-SCH).

### Shared Channel Parameters

The example uses the DL-SCH to describe the processing, which also applies to the UL-SCH.

Select parameters for a transport block transmitted on the downlink shared (DL-SCH) channel.

```
rng(210);           % Set RNG state for repeatability

A = 10000;          % Transport block length, positive integer
rate = 449/1024;    % Target code rate, 0<R<1
rv = 0;             % Redundancy version, 0-3
modulation = 'QPSK'; % Modulation scheme, QPSK, 16QAM, 64QAM, 256QAM
nlayers = 1;        % Number of layers, 1-4 for a transport block
```

Based on the selected transport block length and target coding rate, DL-SCH coding parameters are determined using the `nrDLSCHInfo` function.

```
% DL-SCH coding parameters
cbsInfo = nrDLSCHInfo(A,rate);
disp('DL-SCH coding parameters')
disp(cbsInfo)
```

```
DL-SCH coding parameters
CRC: '24A'
L: 24
BGN: 1
C: 2
Lcb: 24
F: 244
Zc: 240
K: 5280
N: 15840
```

DL-SCH supports multi-codeword transmission (i.e. two transport blocks) while UL-SCH supports only a single codeword. UL-SCH also supports  $\pi/2$ -BPSK modulation in addition to those listed above for DL-SCH.

## Transport Block Processing using LDPC Coding

Data delivered from the MAC layer to the physical layer is termed as a transport block. For the downlink shared channel (DL-SCH), a transport block goes through the processing stages of:

- CRC attachment,
- Code block segmentation and code block CRC attachment,
- Channel coding using LDPC,
- Rate matching and code block concatenation

before being passed on to the physical downlink shared channel (PDSCH) for scrambling, modulation, layer mapping and resource/antenna mapping. Each of these stages is performed by a function as shown next.

```
% Random transport block data generation
in = randi([0 1],A,1,'int8');

% Transport block CRC attachment
tbIn = nrCRCEncode(in,cbsInfo.CRC);

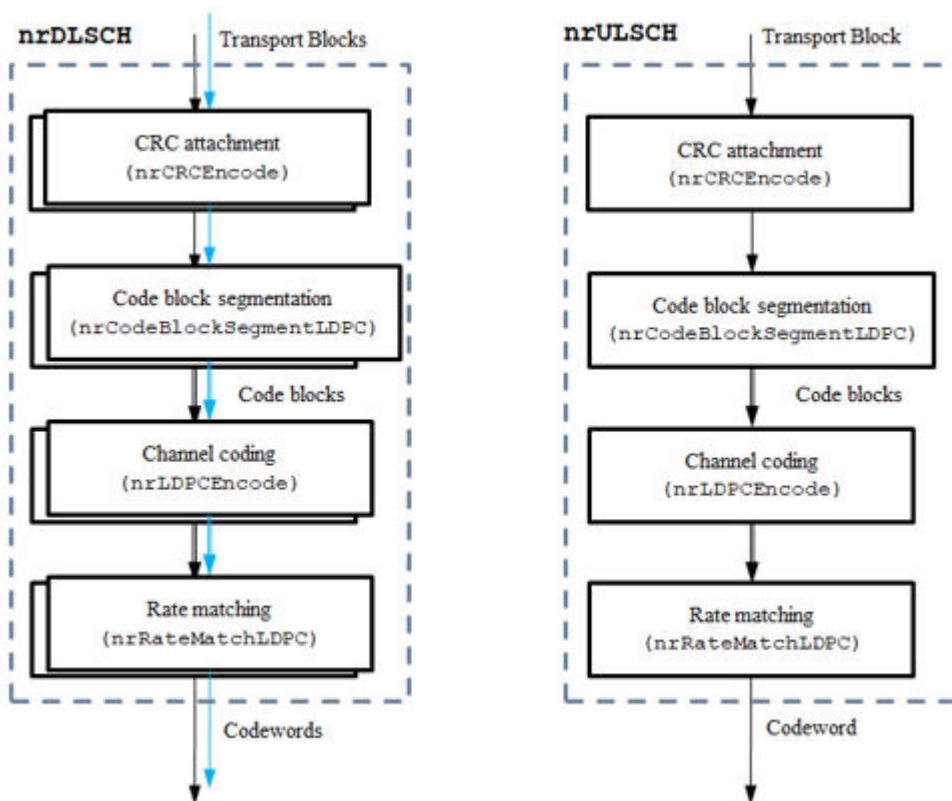
% Code block segmentation and CRC attachment
cbsIn = nrCodeBlockSegmentLDPC(tbIn,cbsInfo.BGN);

% LDPC encoding
enc = nrLDPCEncode(cbsIn,cbsInfo.BGN);

% Rate matching and code block concatenation
outlen = ceil(A/rate);
chIn = nrRateMatchLDPC(enc,outlen,rv,modulation,nlayers);
```

The output number of bits from the rate matching and code block concatenation process must match the bit capacity of the PDSCH, based on the available resources. In this example, as the PDSCH is not modeled, this is set to achieve the target code rate based on the transport block size previously selected.

Similar processing applies for the UL-SCH, where the physical uplink shared channel (PUSCH) is the recipient of the UL-SCH codeword. The following schematics depict the processing for the two channels.



Refer to nrDLSCH and nrULSCH System objects that encapsulate the processing per transport block, with additional support for retransmissions.

### Channel

A simple bipolar channel with no noise is used for this example. With the full PDSCH or PUSCH processing, one can consider fading channels, AWGN and other RF impairments as well.

```
chOut = double(1-2*(chIn));
```

### Receive Processing using LDPC Decoding

The receive end processing for the DL-SCH channel comprises of the corresponding dual operations to the transmit end that include

- Rate recovery
- LDPC decoding
- Code block desegmentation and CRC decoding
- Transport block CRC decoding

Each of these stages is performed by a function as shown next.

```
% Rate recovery
raterec = nrRateRecoverLDPC(chOut,A,rate,rv,modulation,nlayers);

% LDPC decoding
decBits = nrLDPCDecode(raterec,cbsInfo.BGN,25);

% Code block desegmentation and CRC decoding
[blk,blkErr] = nrCodeBlockDesegmentLDPC(decBits,cbsInfo.BGN,A+cbsInfo.L);

disp(['CRC error per code-block: [' num2str(blkErr) ']])

% Transport block CRC decoding
[out,tbErr] = nrCRCDecode(blk,cbsInfo.CRC);

disp(['Transport block CRC error: ' num2str(tbErr)])
disp(['Recovered transport block with no error: ' num2str(isequal(out,in))])

CRC error per code-block: [0 0]
Transport block CRC error: 0
Recovered transport block with no error: 1
```

As the displays indicate, there are no CRC errors at both the code-block and transport block levels. This leads to the transport block being recovered and decoded with no errors, as expected, for a noiseless channel.

Refer to `nrDLSCHDecoder` and `nrULSCHDecoder` System objects that encapsulate the receive processing per codeword, with additional soft-combining of retransmissions for improved performance.

## See Also

### System Objects

`nrDLSCH` | `nrDLSCHDecoder` | `nrULSCH` | `nrULSCHDecoder`

### Functions

nrCRCDecode | nrCRCEncode | nrCodeBlockDesegmentLDPC |  
nrCodeBlockSegmentLDPC | nrDLSCHInfo | nrLDPCDecode | nrLDPCEncode |  
nrRateMatchLDPC | nrRateRecoverLDPC | nrULSCHInfo

### More About

- “NR PDSCH Throughput”
- “NR PUSCH Throughput”